

## A Causal Microwave Circuit Theory and Its Implications

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**Abstract- We will describe a new causal power-normalized waveguide equivalent-circuit theory and explore its implications. The new theory marries a power normalization with additional constraints that enforce simultaneity of the theory's voltages and currents and the actual fields in the circuit.**

The causal waveguide circuit theory of [1] requires that  $Z_0(\omega)$ , the characteristic impedance of the guide, be minimum phase. That is, the theory requires that  $\hat{Z}_0(t) = 0$  and  $\hat{Y}_0(t) = 0$  for  $t < 0$ , where  $\hat{Z}_0(t)$  and  $\hat{Y}_0(t)$  are the inverse Fourier transforms of  $Z_0(\omega)$  and  $Y_0(\omega) \equiv 1/Z_0(\omega)$ . This condition is required to ensure that the waveguide responds to inputs after, not before, it is excited.

The voltage  $v$  and current  $i$  in [1] are defined by

$$\mathbf{E}_t(\mathbf{r}, z) = \frac{v(z)}{v_0} \mathbf{e}_t(\mathbf{r}); \quad \mathbf{H}_t(\mathbf{r}, z) = \frac{i(z)}{i_0} \mathbf{h}_t(\mathbf{r}), \quad (1)$$

where  $\mathbf{r} = (x, y)$  is the transverse coordinate,  $\mathbf{E}_t$  and  $\mathbf{H}_t$  are the total electric and magnetic fields in the guide,  $\mathbf{e}_t$  and  $\mathbf{h}_t$  are the modal electric and magnetic fields of the single propagating mode, and  $v_0$  and  $i_0$  are normalization factors.

The power normalization is achieved by imposing the constraint [2]

$$v_0 i_0^* = p_0 \equiv \int_S \mathbf{e}_t \times \mathbf{h}_t^* \cdot \mathbf{z} dS, \quad (2)$$

which ensures that the time-average power is  $p = \frac{1}{2} v i^*$ . The characteristic impedance  $Z_0$  of the waveguide is defined by

$$Z_0 \equiv \left. \frac{v}{i} \right|_{c_-=0} = \frac{v_0}{i_0} = \frac{|v_0|^2}{p_0^*} = \frac{p_0}{|i_0|^2}, \quad (3)$$

which shows that the power normalization sets the

phase of  $Z_0$  equal to the phase of  $p_0$ , a fixed property of the guide.

However, the minimum phase constraint imposed by causality is a strong one [3], [4]. It requires that the Hilbert transform of  $\ln|\lambda Z_0|$ , where  $\lambda$  is a constant, be equal to  $\arg(Z_0) = \arg(p_0)$ , a fixed property of the guide. As a result, the characteristic impedance of the guide is determined within a constant.

The implications of the new causal circuit theory of [1] are significant. The power flow  $p_0$  is real in a lossless coaxial transmission line, so the phase of  $Z_0$  is 0, and in the causal circuit theory  $|Z_0|$  must be constant. We also show that the characteristic impedance of a rectangular waveguide must be proportional to the wave impedance of the guide: the choice  $|Z_0| = 1$ , allowed in conventional waveguide circuit theories, is not admissible in the causal theory.

### REFERENCES

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