

# Antenna-Coupled TES Array Development in Astrophysical Applications

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# Overview

- Instrumentation requirements in Cosmic Microwave Background (CMB) polarization anisotropy
- Cryogenic detectors
- Antenna-coupled TES array architecture and characterization at Caltech/JPL
- TES array multiplexing and systematic issues
- Cryogenic detectors development at Argonne Lab
- Summary

# Tools in Understanding the Universe

- Theories
  - Big Bang
  - Inflation
  - Baryogenesis
  - Nucleosynthesis
  - Recombination
  - Star formation, then galaxies, clusters of galaxies, ...
- Tests
  - Hubble's law
  - Nucleosynthesis data
  - Structure formation history
  - **Cosmic Microwave Background (CMB)**

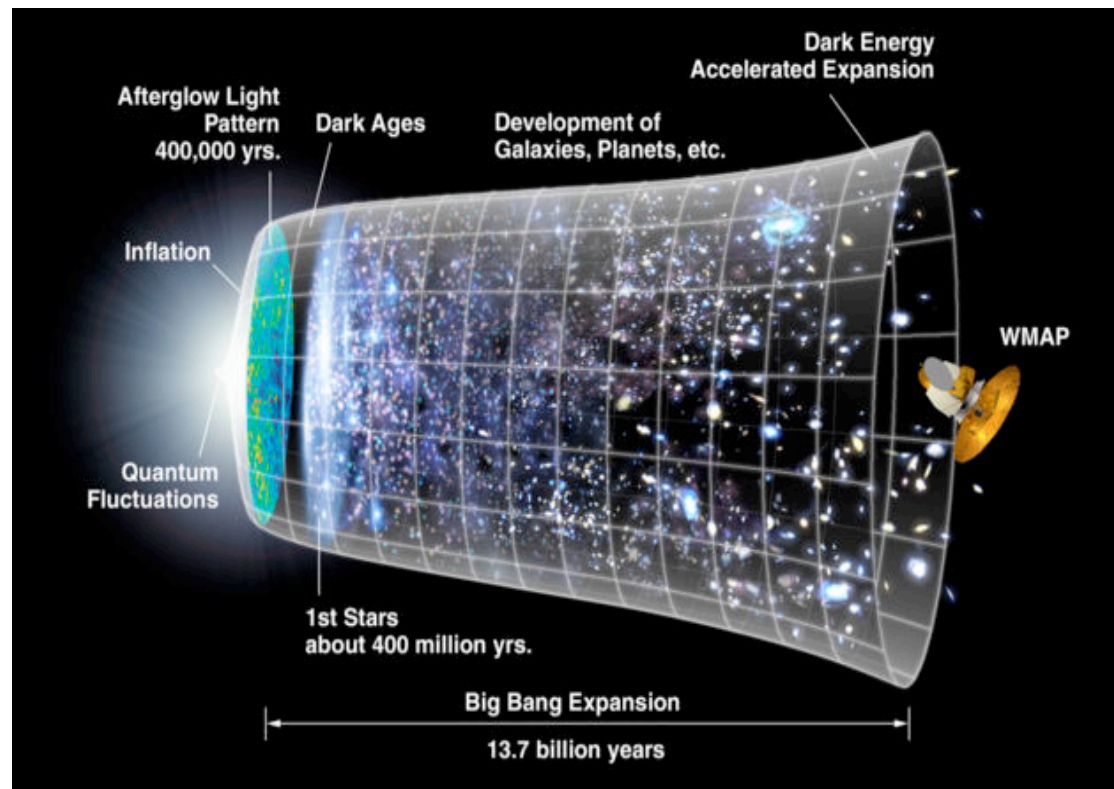
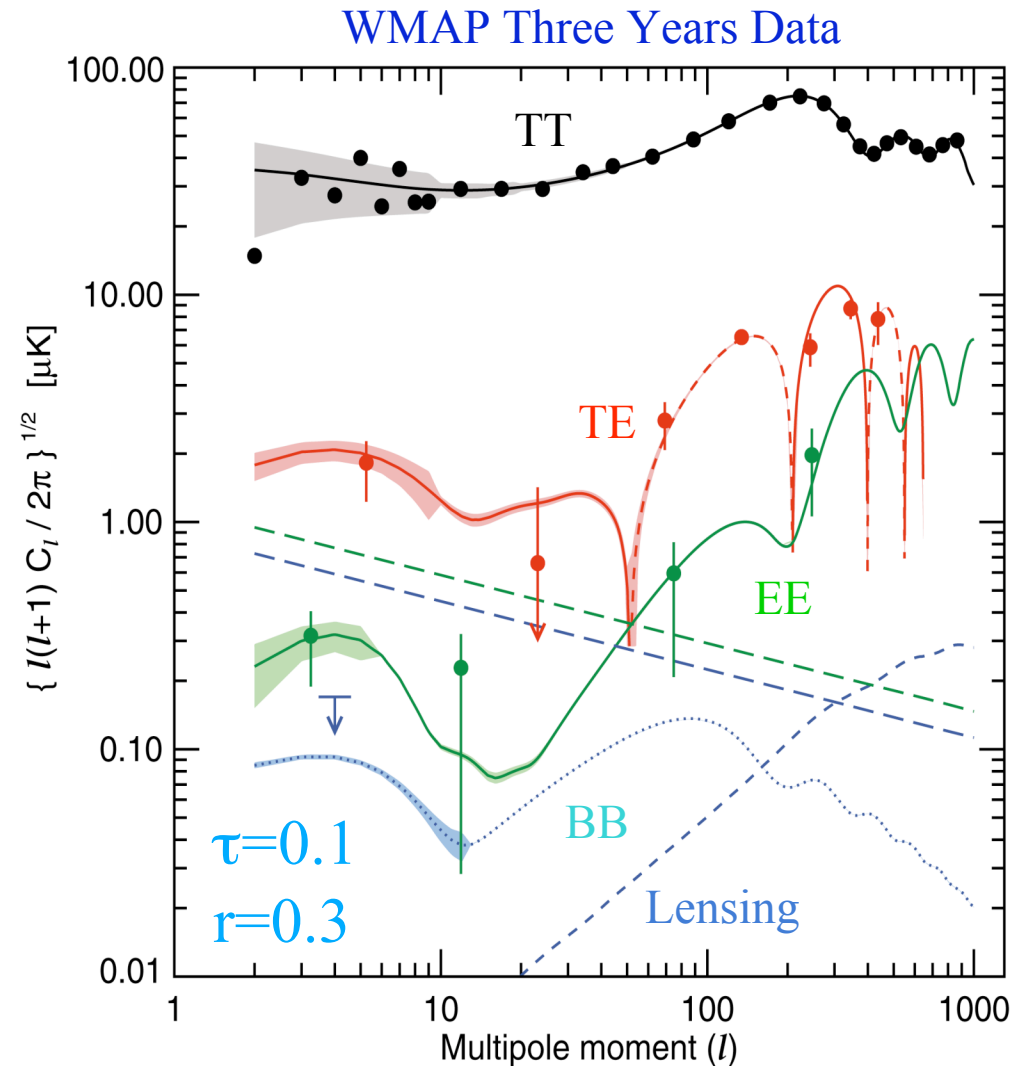


Image from WMAP press release

# Status of CMB Observations

- CMB Blackbody Temperature
  - ☑ Observed
  - ☑ Well Characterized
- CMB Temperature Anisotropy
  - ☑ Observed
  - ☑ Well Characterized
- CMB E-mode Polarization
  - ☑ Observed (DASI, B03, CBI, CAPMAP, QUAD, WMAP)
  - Not Well Characterized
- CMB B-mode Polarization
  - Not observed, Possible by Planck, SPIDER, and SPTpol in future
  - Not Characterized



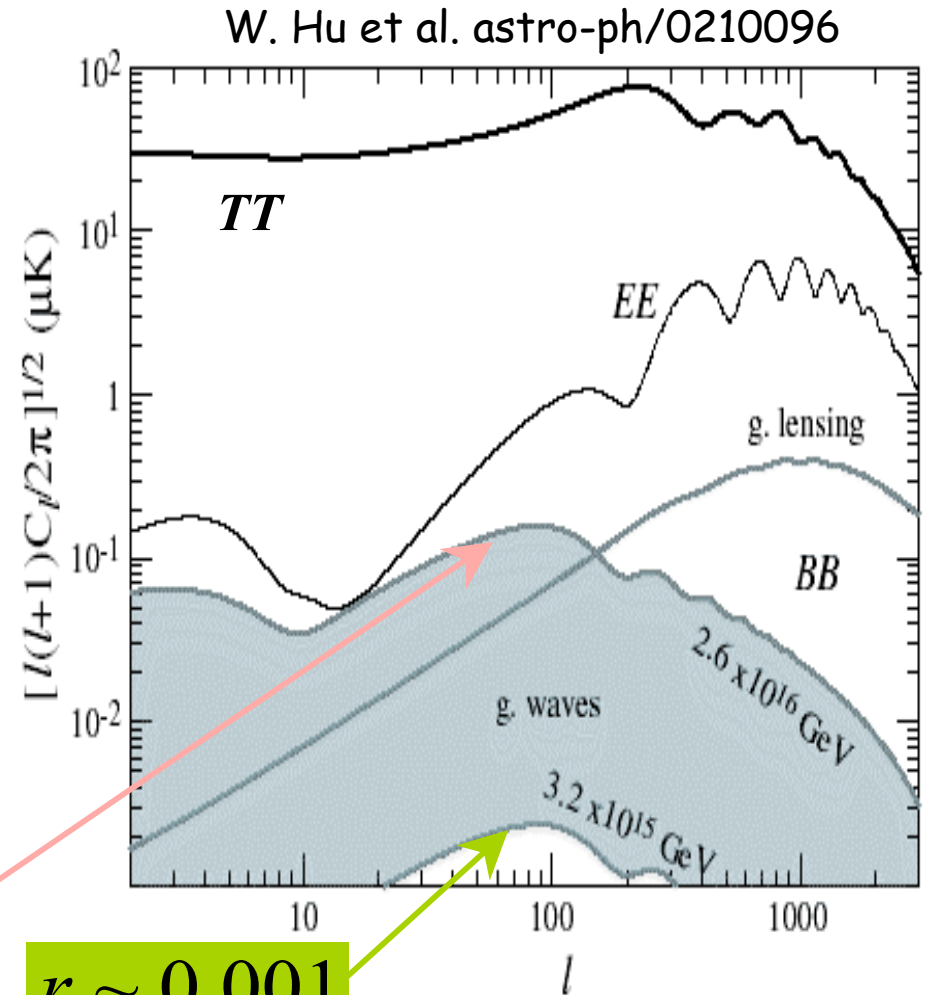
# Why CMB B-mode Polarization Measurement?

- The fingerprint at the time of inflation ( $10^{-35}$  s) is encoded on the **cosmic gravitational-wave background (CGB)**
- Exploration and test of **high energy physics in GUT**
- CGB stretches and compresses spacetime on the surface of last scattering -- transmitting this information as a curl component in the polarization

$$r \equiv \frac{T^2}{S^2} \propto \left( \frac{E_I}{M_{pl}} \right)^4$$

$r \sim 0.1$

$r \sim 0.001$

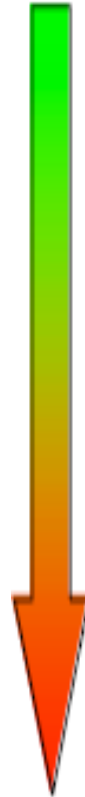


# Detector Sensitivity for CMB B-Modes

- CMB temperature
- CMB anisotropy angular power spectrum
- CMB polarization anisotropy angular power spectra

– E-modes

– B-modes



Increasing  
difficulty

$$\langle T \rangle = 2.725K$$

$$\Delta T_{rms} \sim 20\mu K$$

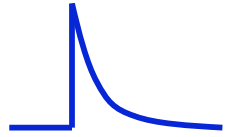
$$\Delta T_{E,rms} \sim 2\mu K$$

$$\Delta T_{B,rms} \leq 0.1\mu K$$

# CMB Detectors

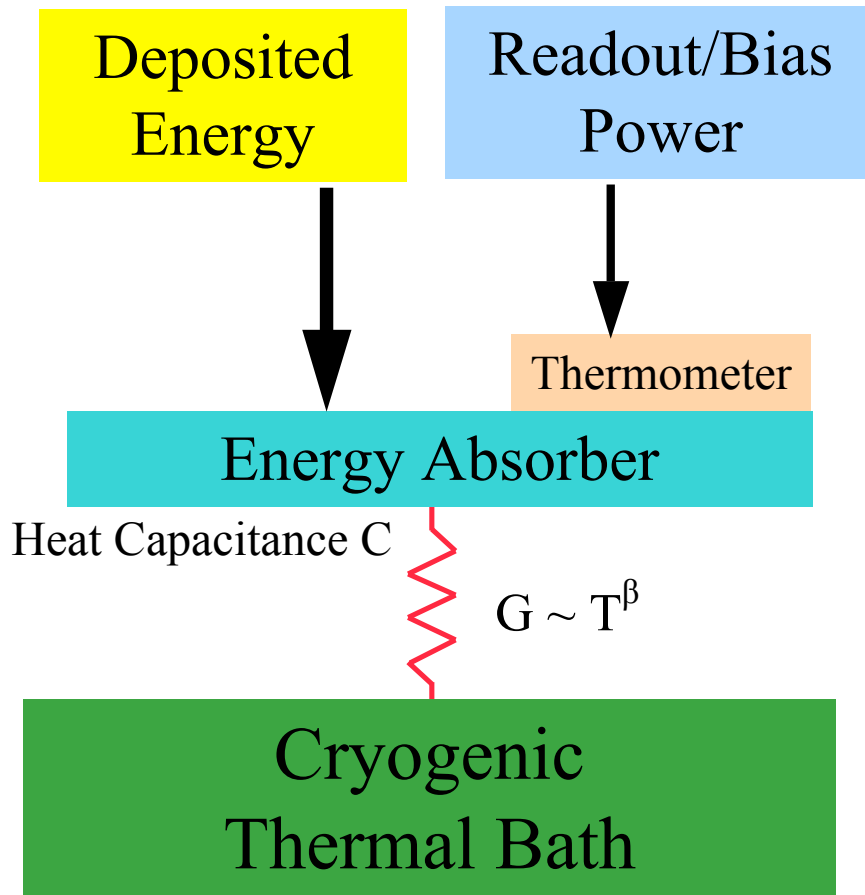
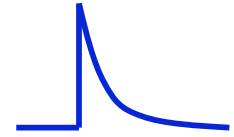
- Interferometric
  - measuring microwave electric field amplitude and phase at the same time
  - n-antennas (or telescopes), number of correlated channels is proportional to  $n^2$
  - Moderate noise level  $\sim 1 \text{ mK Hz}^{-0.5}$  for each antenna, and instrument sensitivity inversely scales with n
  - 4 ~ 20 K
- Coherent
  - Feed horns or waveguides, then OMT
  - Noise level is moderate: WMAP,  $0.65 \text{ mK Hz}^{-0.5}$  at K;  $1.48 \text{ mK Hz}^{-0.5}$  at W
  - 4 ~ 20 K
- Bolometric
  - Measuring microwave electric field amplitude
  - Feed horns or wave guides could be removed
  - Noise level is low,  $0.1 \sim 0.2 \text{ mKHz}^{-0.5}$ , as low as  $0.03 \text{ mKHz}^{-0.5}$  in space operated below 0.1K
  - Large format receivers array with a single telescope
  - Below 1 K, cryogenic detectors

Temperature



# Cryogenic Detector Model

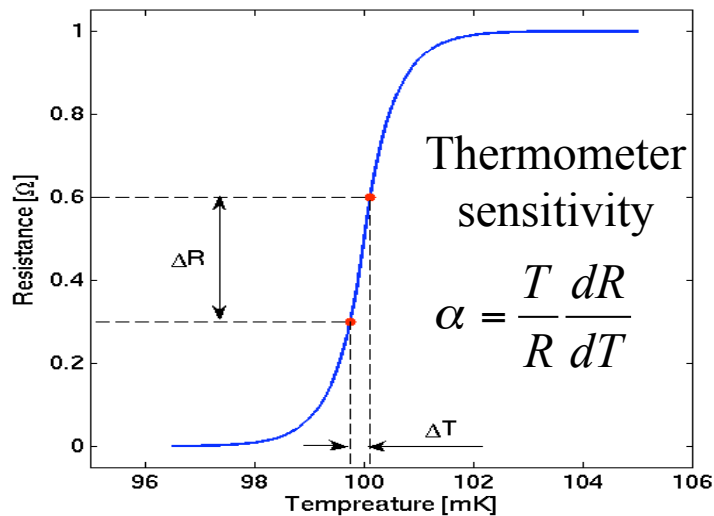
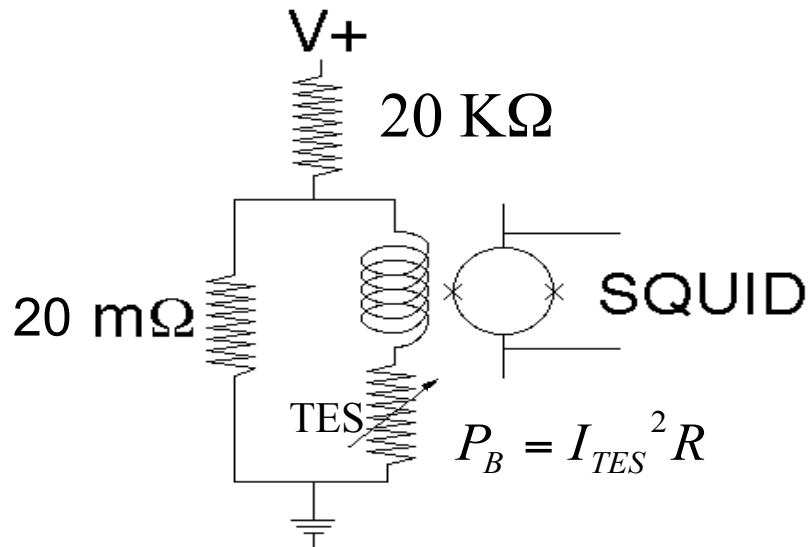
Readout



- Energy coupling**
  - Antenna and/or microstrip
  - High Z metal
  - Semiconductor crystal
- Thermal design**
  - Operated below 1K
  - Absorber mass depends on energy to be measured
  - Thermal link: direct substrate contact or connection with reduced thermal conductance
- Thermometer**
  - Resistive: NTD, TES
  - Junctions: SIS, NIS
  - Magnetic: electron spin
  - Inductance: MKIDs



# TES Detectors



- Conventional superconductor film
- Voltage bias with a negative feedback
- Temperature change is converted to a current change

$$C \frac{dT}{dt} = P_B - G(T - T_b) + Q\delta(t)$$

$$\Delta I \approx -\frac{Q}{C} \frac{\alpha \cdot I_{TES}}{T} \exp\left(-\frac{t}{\tau}\right)$$

$$\tau \approx \frac{C}{G} \frac{1}{1+g} \quad g = \frac{P_B \alpha}{GT}$$

$$S(\omega) \approx \frac{1}{V_B} \frac{1}{1+i\omega\tau}$$

# Why TES Detectors?

- High performance thermometer
  - Sharp superconducting transition, good sensitivity
  - Voltage bias (electro-thermal feedback)
    - linear broad band response: up to a few tens kHz
    - Fast response time: a few tens  $\mu\text{s}$  ~ a few tens ms
    - Modest variations in  $T_c$  across wafer are accommodated
  - Low dissipation: TES + shunt resistor  $\sim 0.1\text{nW}$
  - Low noise  $NEP \sim 10^{-18}\text{W} / \sqrt{\text{Hz}}$
- Low impedance devices
  - Immune to microphonic noise
  - Impedance matching to SQUID amplifier's dynamic resistance
- Applications
  - Calorimeter and micro-calorimeters: particle detectors, for example **CDMS**; improved energy dispersive spectrometer in material science; constellation X-ray in astrophysics; radiation remote sensing
  - **Bolometers: UV, visible, infrared, and millimeter photons**

# Review of Noise in CMB Measurement

- Noise specified as **Noise Equivalent Power** (NEP), power incident on detector that can be detected at  $1\sigma$  in 1 sec, units of  $\text{W Hz}^{-0.5}$ 
  - detector noise: phonon noise, Johnson noise of resistors, amplifier noise.
  - **Background Limited Infrared Photodetector** (BLIP) noise: shot noise on DC optical load
  - sky noise: variations in sky loading
- These yield **Noise Equivalent Flux Density** (NEFD): flux density (Jy) that can be detected at  $1\sigma$  in 1 sec, units of  $\text{Jy Hz}^{-0.5}$ 

$$\text{NEFD} = \frac{\text{NEP}}{\eta A \Delta \nu}$$
- Beam size defines **Noise Equivalent Surface Brightness** (NESB), units of  $(\text{Jy}/\text{arcmin}^2) \text{ Hz}^{-0.5}$ 

$$\text{NESB} = \frac{\text{NEFD}}{\Omega_{\text{beam}}}$$
- Can then calculate **Noise Equivalent Temperature** ( $\text{NET}_{\text{CMB}}$ ), units of  $(\mu\text{K}_{\text{CMB}}/\text{beam}) \text{ Hz}^{-0.5}$ 

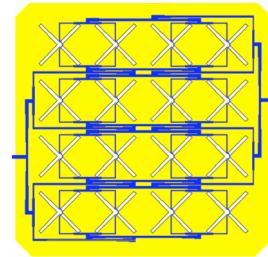
$$\text{NET}_{\text{CMB}} = \text{NESB} \left( \left. \frac{dI_\nu}{dT} \right|_{T_{\text{CMB}}} \right)^{-1}$$
- For diffraction limited optics  
And at Rayleigh – Jeans limit
 
$$\text{NET}_{\text{CMB}} \approx \frac{\text{NEP}}{2k_B \eta \Delta \nu}$$

# Need Large Bolometer Array

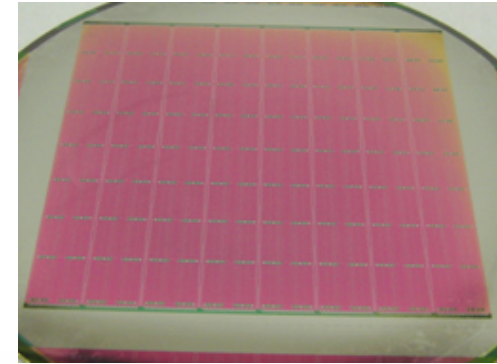
- TES bolometers are operated at 100~300mK
- Current fully instrumented bolometer NEP at  $1e-18 \text{ W Hz}^{-0.5}$
- Photon noise NEP at about  $5\sim 8e-18 \text{ W Hz}^{-0.5}$  in space, and about  $1e-17 \text{ W Hz}^{-0.5}$  near ground
- These can be transferred to CMB measurement sensitivity of  $50\sim 200 \mu\text{K}_{\text{CMB}} \text{ Hz}^{-0.5}$ , which depends on bolometers operation temperature, instrumentation configuration, and frequencies
- To reach  $1 \mu\text{K}_{\text{CMB}} \text{ Hz}^{-0.5}$  for B-mode CMB detection, 1000 or more bolometers are needed at each specific frequency band
- Compact photolithographic antenna-coupled TES bolometer large array

# Large Detector Array

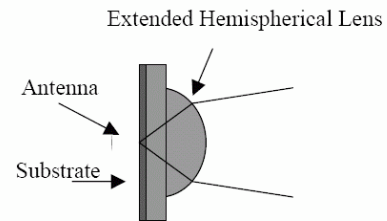
**Antenna arrays**  
(Caltech/JPL)



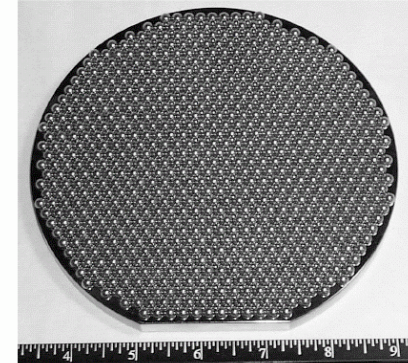
Spider/SPUD



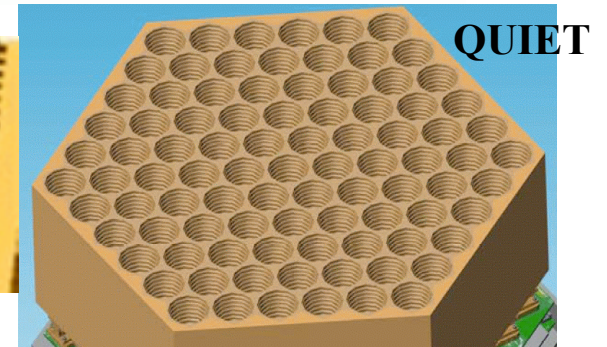
**Lenslets (Berkeley)**



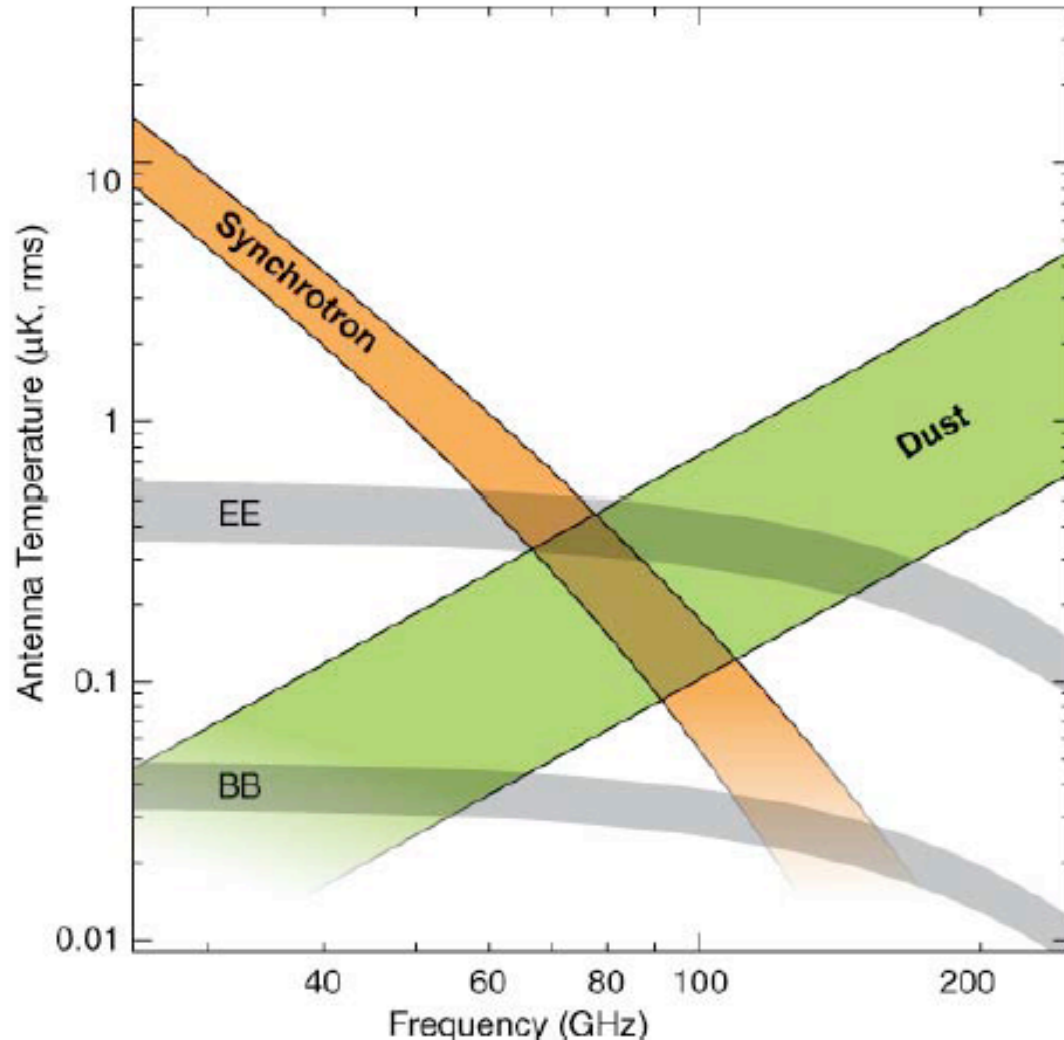
Polarbear



**Platelets (U of C)**



# Multi – bands for Foreground Subtraction



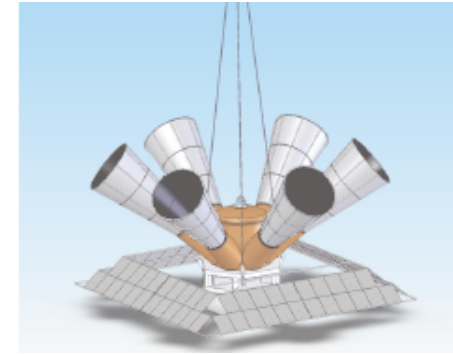
Task Force Report  
on CMB Research

For  $r = 0.01$ , more  
than 10 fold  
background  
reduction needed

Need multi-frequency  
bands for foreground  
subtraction

# SPIDER: Balloon Borne Flight

CMB polarization B-mode measurement  
pathfinder: a large angular scale millimeter  
wave polarimeter with antenna-coupled TES  
large format arrays suitable for multiplexing



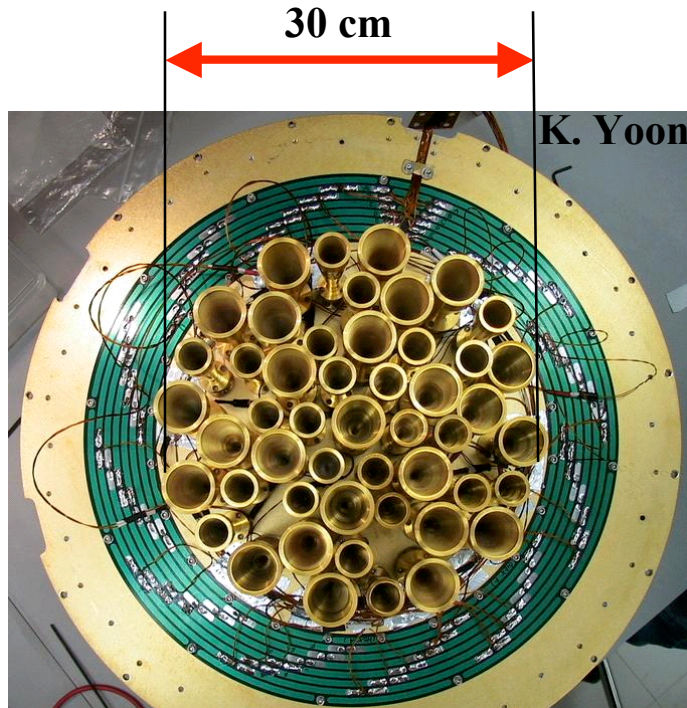
Observing Band (GHz)	Band width (GHz)	Beam FWHM (arcmin)	Number of Spatial Pixels	Number of Detectors	Single Detector Sensitivity ( $\mu\text{K}_{\text{CMB}} \text{Hz}^{-0.5}$ )	Instrument Sensitivity ( $\mu\text{K}_{\text{CMB}} \text{Hz}^{-0.5}$ )
80	19	72	100	200	110	7.8
100	24	58	2×144	2×288	100	4.2
145	35	40	256	512	100	4.4
225	54	26	256	512	204	9.0
275	66	21	256	512	351	15.5

Assume the overall optical efficiency 50%. Operated at 250mK

# Format Receiver Arrays

## *The state-of-the-art:*

- Discrete elements: feeds, filters, detectors
- Massive and expensive, hand-assembled
- Low packing efficiency
- Individual JFET readout with moderate noise

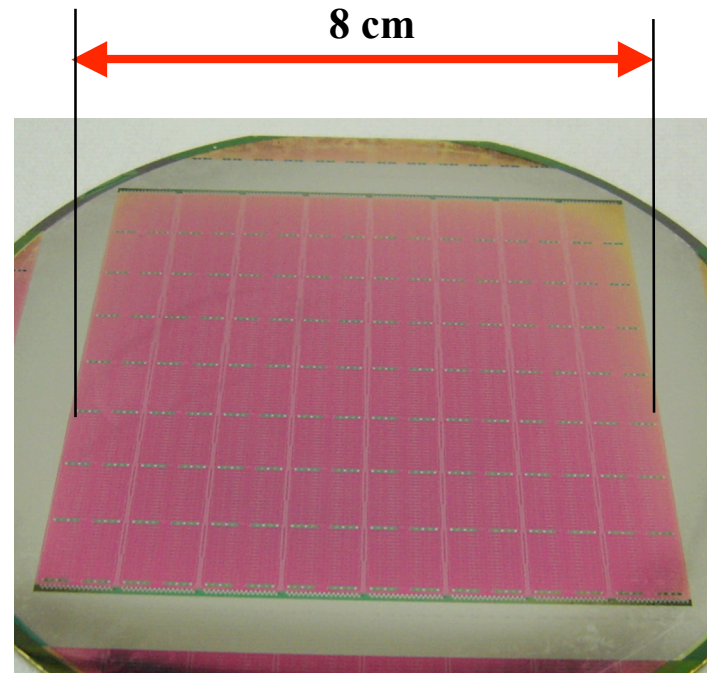


**BICEP focal plane (98 detectors)**

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## *The future:*

- Integrating all components on a Si wafer
- Compact, inexpensive, photolithographic
- High packing density
- TES low impedance enables SQUID multiplexed read-out with low noise



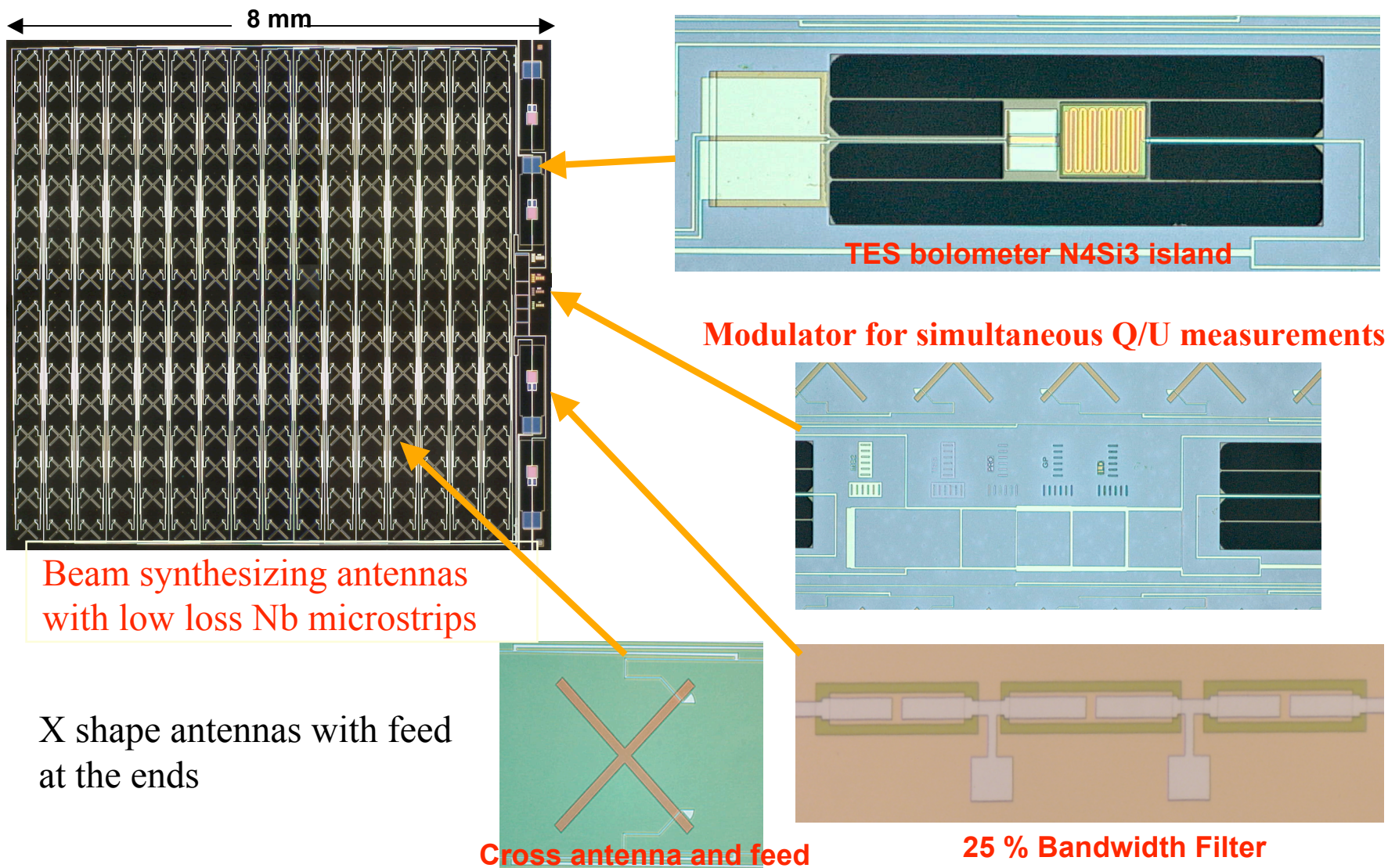
**Antenna-coupled TES array  
(256 Q/U detectors)**

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# 150 GHz Dual Polarization Antenna Coupled TES

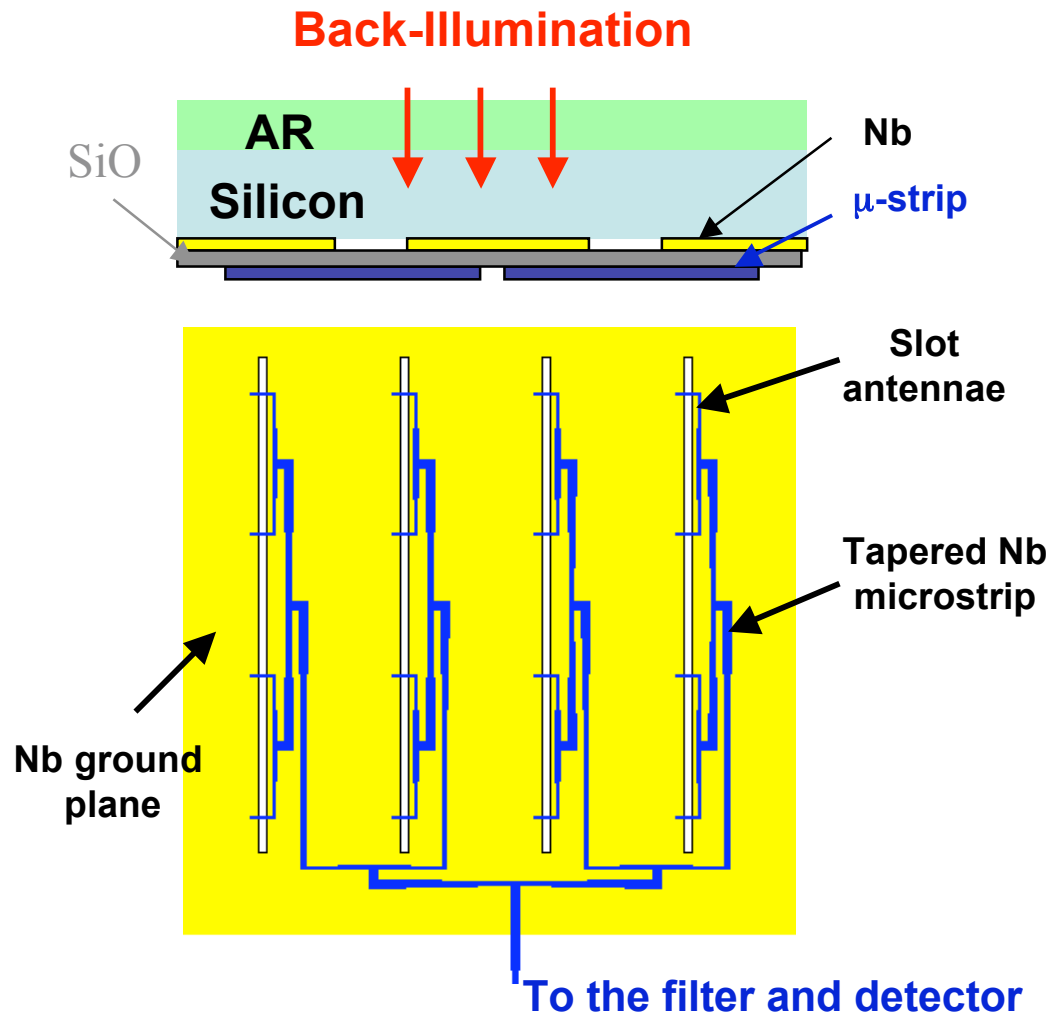


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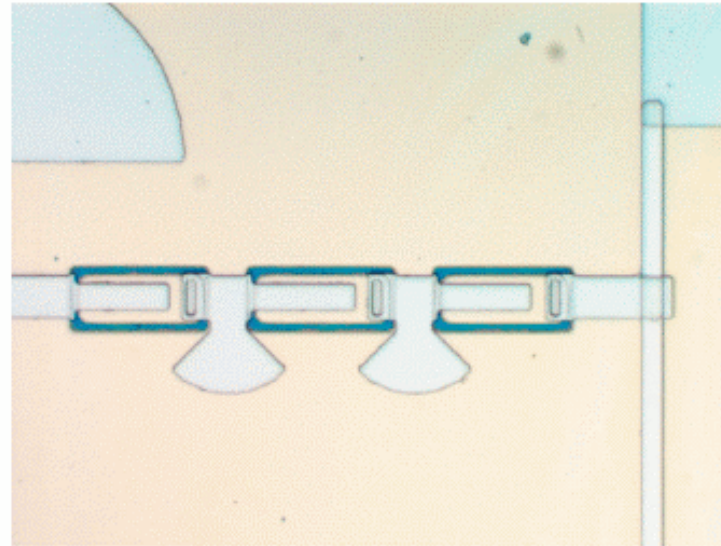
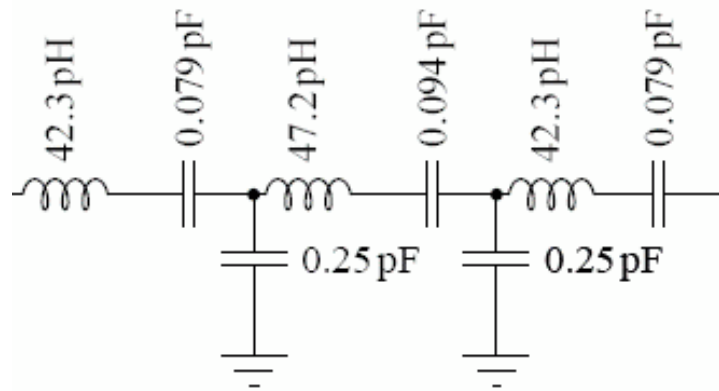
# Single Polarization Antenna Architecture



- Long slot antennas on the Nb ground plane provide beam collimation
- low loss Nb microstrip is coupled to slot antenna with stubs or vias
- Tapered Nb binary sum tree transport microwave to filters and detectors
- 50~500 GHz achievable
- Evaporated dielectric SiO<sub>2</sub>
- Compact planar design with contact microfabrication

# Filter

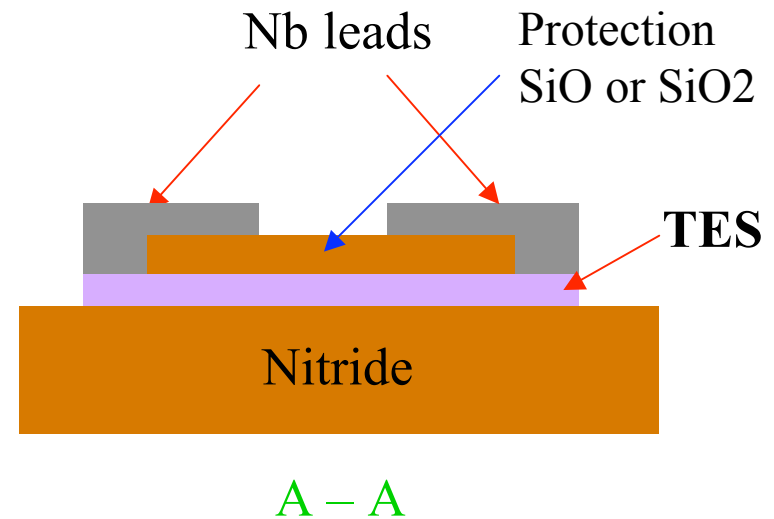
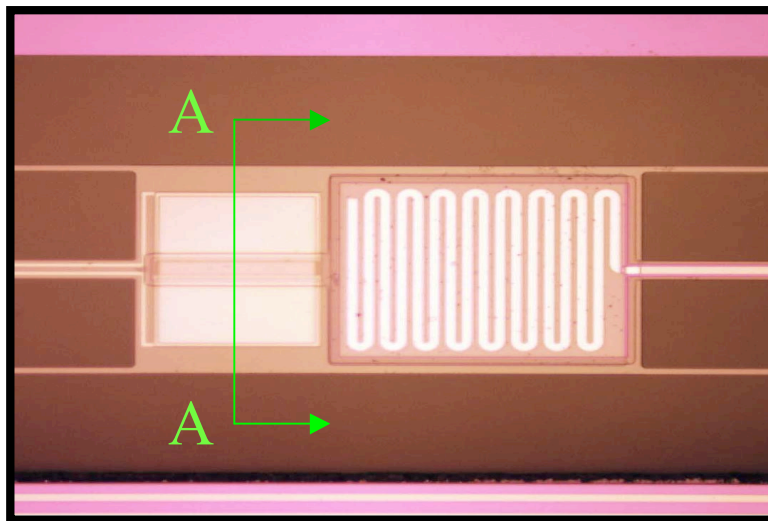
- Lithographic LC filter defines the band pass
- Nb (ground plane) – SiO – Nb (microstrip elements) structure
- 25% Bandwidth



**Designed by A. Goldin**

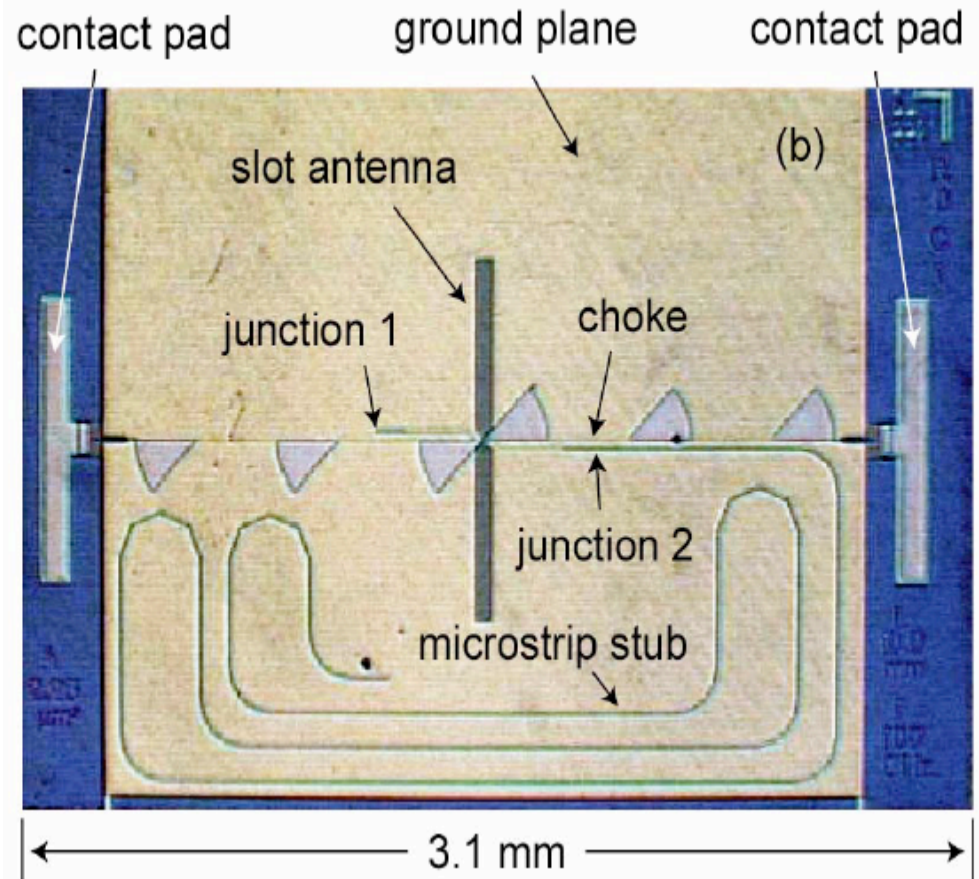
# Bolometer

- Bolometer released by deep reactive ion etch
- Nitride legs support nitride membrane island
- Lossy metal microstrip as termination – wide bandwidth
- Superconducting transition edge sensor
- Quick thermalization time, fast bolometer relaxation time constant



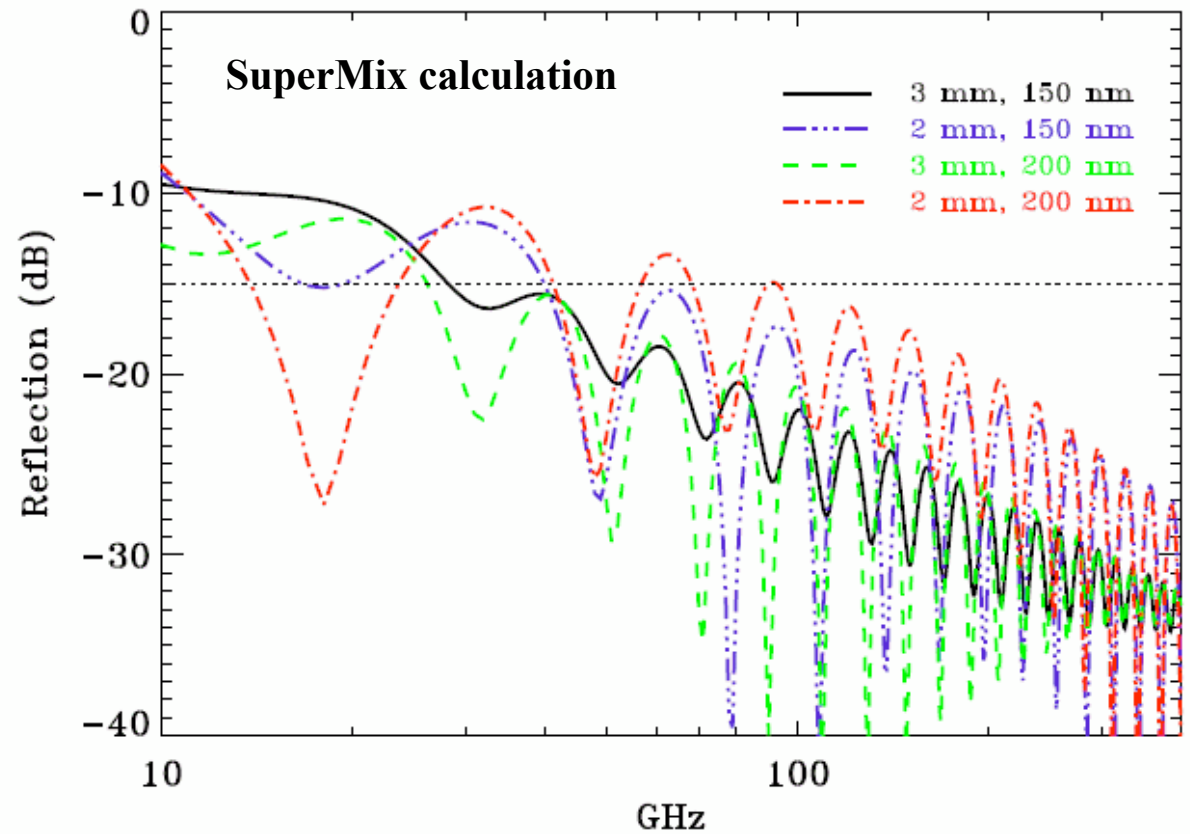
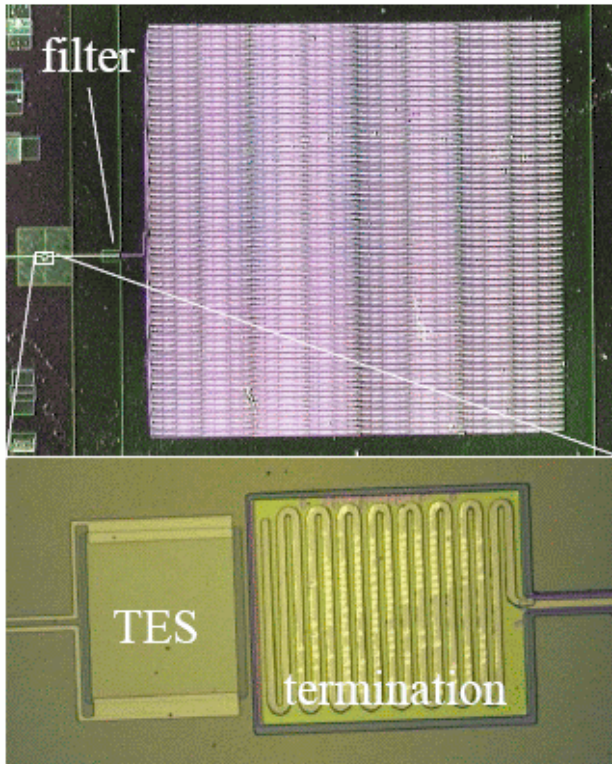
# Microstrip Loss

- Tested with SIS junctions at 4.2K and 1.5K
- The loss tan is  $1.2 \pm 0.3 \times 10^{-3}$
- The amplitude 1/e attenuation length is 30~40cm, corresponding about 12% loss for 100 GHz detector
- The main loss comes from dielectric (two level systems)



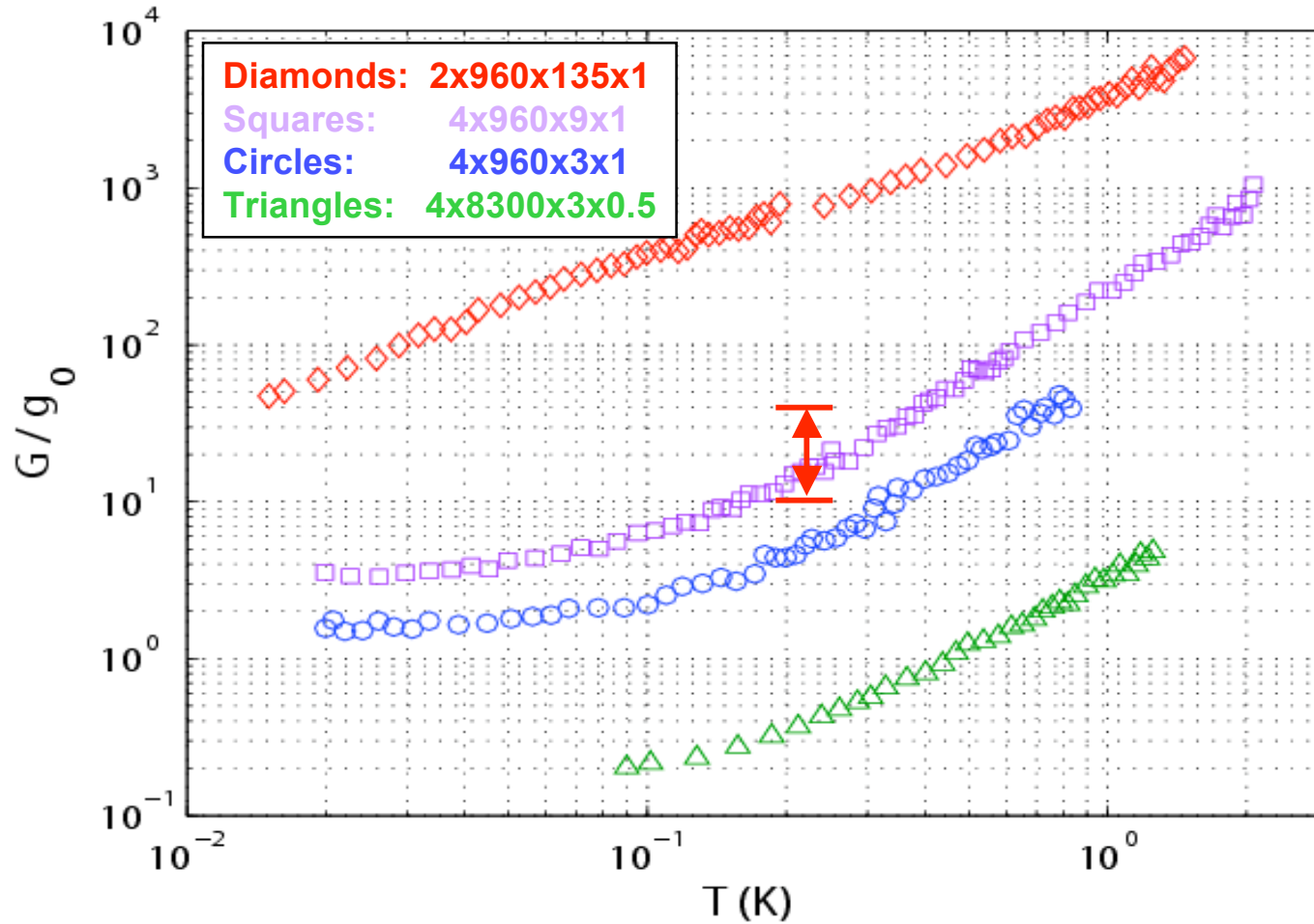
100 GHz test chip with 11.4 mm long Nb/SiO/Nb microstrip stub. [A. Vayonakis](#)

# Wide Band Termination



- The meandering absorbing microstrip has  $> 99\%$  absorbing efficiency over 30% bandwidth
- Very tolerant to fabrication variations, e.g., thickness or resistivity

# Nitride Legs



$g_0 = \pi^2 k_b^2 T / 3h \sim 1 \text{pW/K}^2 \times T$       **SPIDER  $G/g_0 = 10\text{-}50$  at 250 mK**

# Microfabrication

- Metallization
  - UHV sputter deposition systems
  - Current materials set: Nb, Al, Ta, Mo, Ti, Cu, Au, NbN, NbTiN
- Dielectric
  - LPCVD – low stress silicon nitride
  - Sputtered SiO<sub>2</sub> with substrate bias and O<sub>2</sub> compensation.
  - Evaporated SiO.
- Lithography
  - Stepper  $\lambda=248\text{nm}$ ,  $0.25\mu\text{m}$  resolution
  - Contact lithography:  $\lambda=436, 365, 320\text{ nm}$
- Etch
  - 1 Chlorine chemistry, 1 Fluorine chemistry – 75mm, 100mm, 150mm wafers
  - STS deep trench reactive ion etcher - DRIE

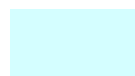


# Multi-layer Structure Processing

	Layer	Etch/Lift-off	Material	Thickness
1.	Support legs pre		$\text{Si}_3\text{N}_4$	1000nm
2.	Ti-TES	$\text{CCl}_2\text{F}_2/\text{O}_2$ etch	Ti	250nm
3.	Protect	$\text{CHF}_3/\text{O}_2$ etch	SiO or $\text{SiO}_2$	150nm
4.	Ground Plane	$\text{CCl}_2\text{F}_2/\text{O}_2$ etch	Nb	150nm
5.	ILD	$\text{CHF}_3/\text{O}_2$ etch	SiO or $\text{SiO}_2$	300nm
6.	Resistor	Lift-off	Au	150nm
7.	Resistor Protect	Lift-off	SiO	150nm
8.	Microstrip	$\text{CCl}_2\text{F}_2/\text{O}_2$ etch	Nb	400nm
9.	FSN	$\text{BCl}_3/\text{Cl}_2$ etch	Al	500nm

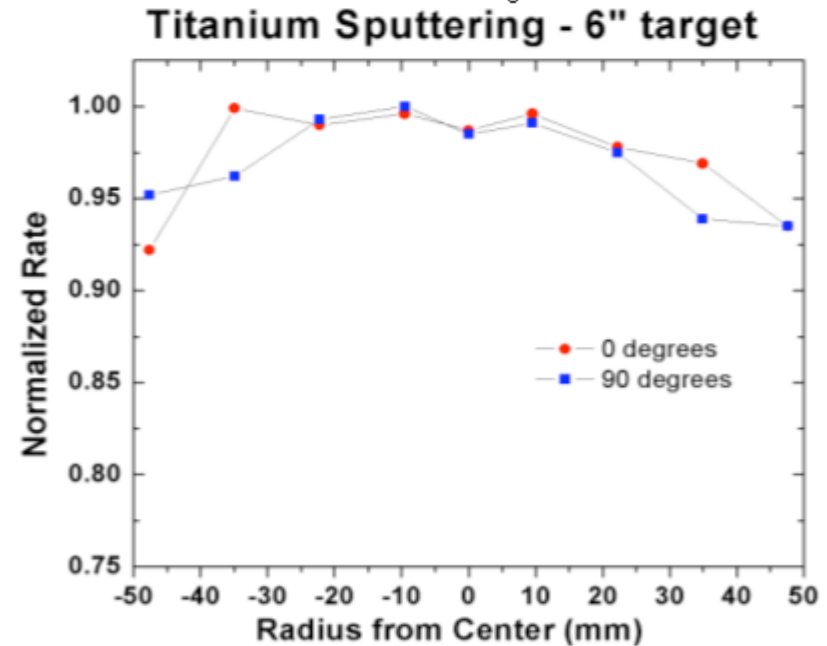
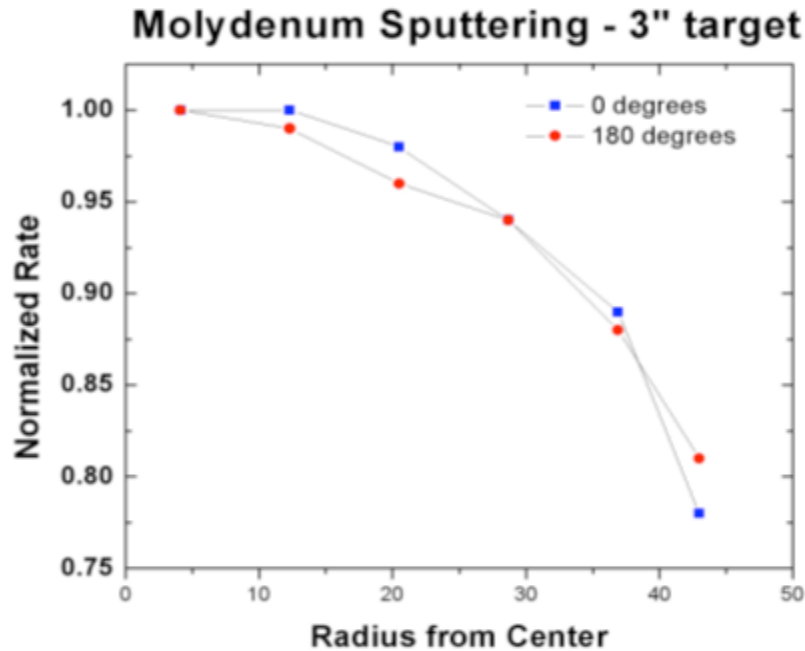


Metallization



Dielectric

# TES Uniformity

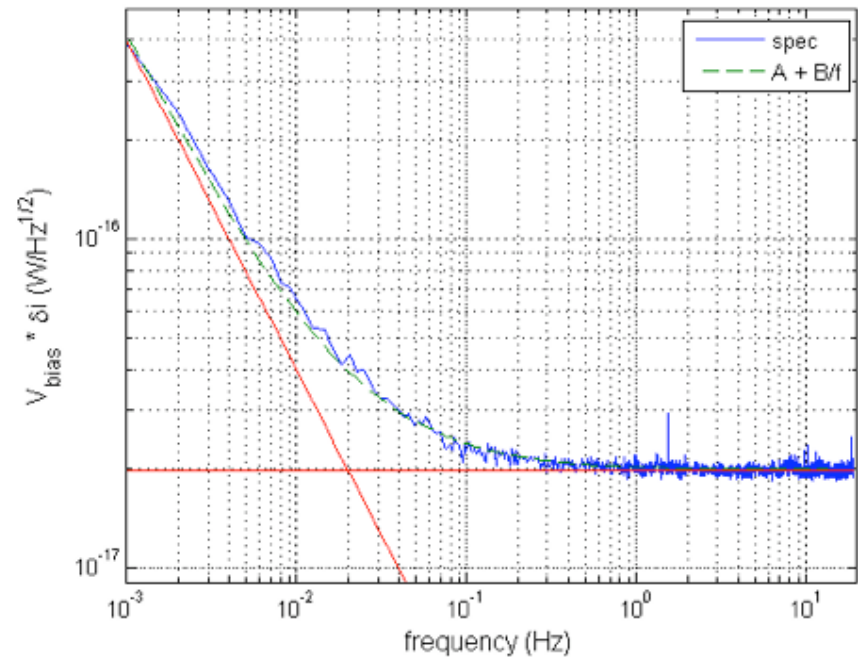
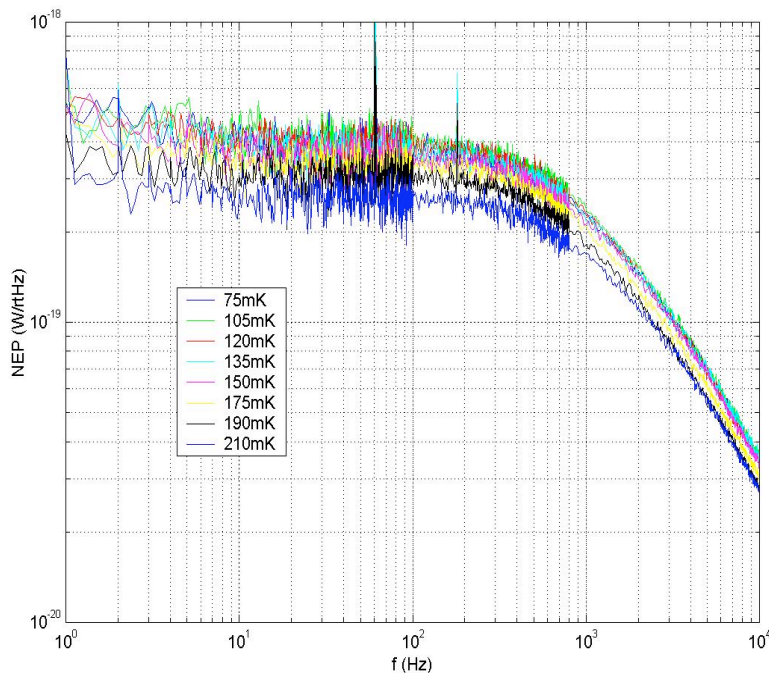


- TES transition temperature could change dramatically with metal film thickness and the contents of impurity contaminations
- Increase sputtering gun (target) radius, could control the superconducting film thickness uniformity with other sputtering parameters tuning
- Devoted facility to avoid impurity contaminations
- The target is to control  $T_c$  within 95% for Ti

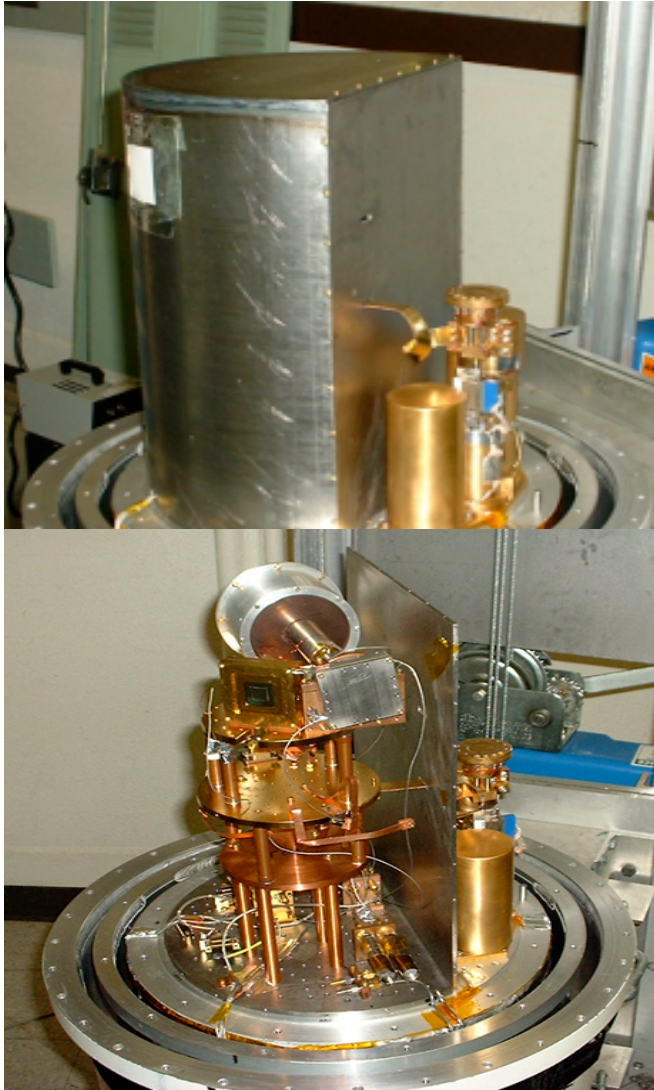
# Detectors Performance (Dark Tests)

- Mo/Au bilayer,  $T_c \sim 220$  mK
- $G \sim 1$  pW/K
- $NEP \sim 3 \times 10^{-19}$  W Hz $^{-1/2} \propto \sqrt{4k_B G T^2}$
- $1/f$  knee  $< 1$  Hz
- Rolling off frequency 800Hz  $\propto G/C$

- Ti/Al dual transitions,  $T_c \sim 450$  mK
- $G \sim 50$  pW/K
- $NEP \sim 2 \times 10^{-17}$  W Hz $^{-1/2}$
- $1/f$  knee  $\sim 50$  mHz
- Rolling off frequency  $> 1000$ Hz



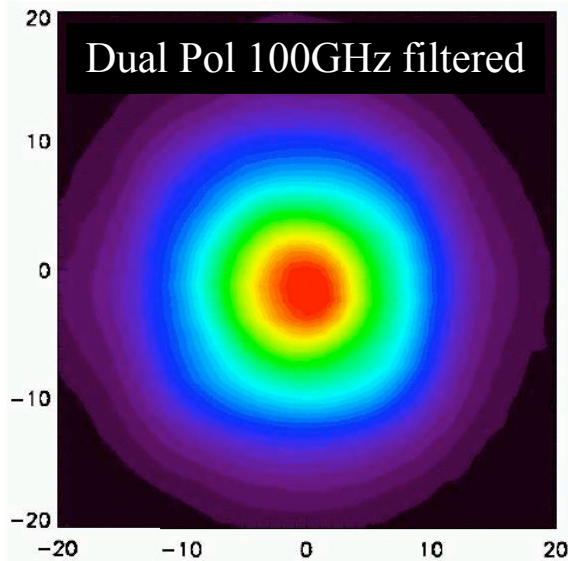
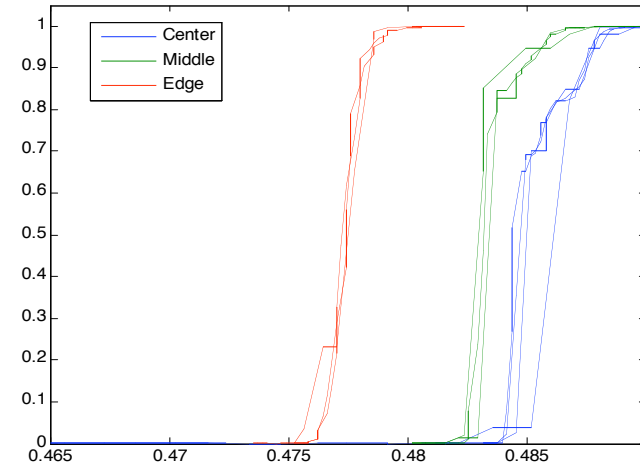
# Antenna-Coupled TES Test facility



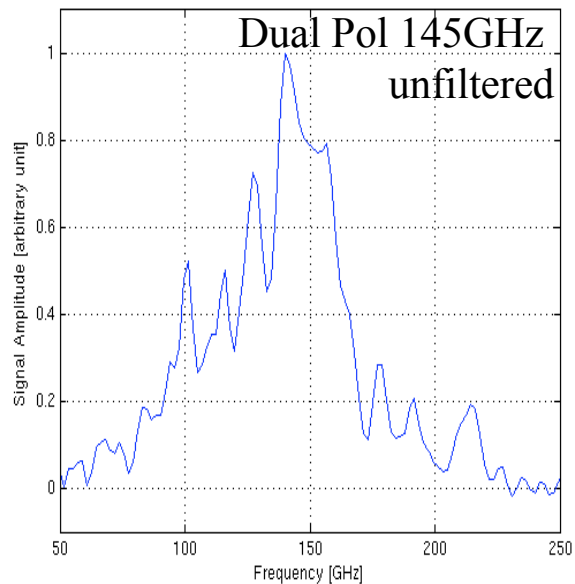
- 3 stage He absorption refrigerator with base temperature of 220mK
- Niobium shielding
- Optical windows
- Internal black body for optical efficiency test
- AC resistance bridges for thermometry
- PID temperature control
- Six channels of quantum physics SQUIDs
- Lock-ins: Stanford research
- Fourier transformation spectrometer
- Home made beam mapper

# Test Tasks

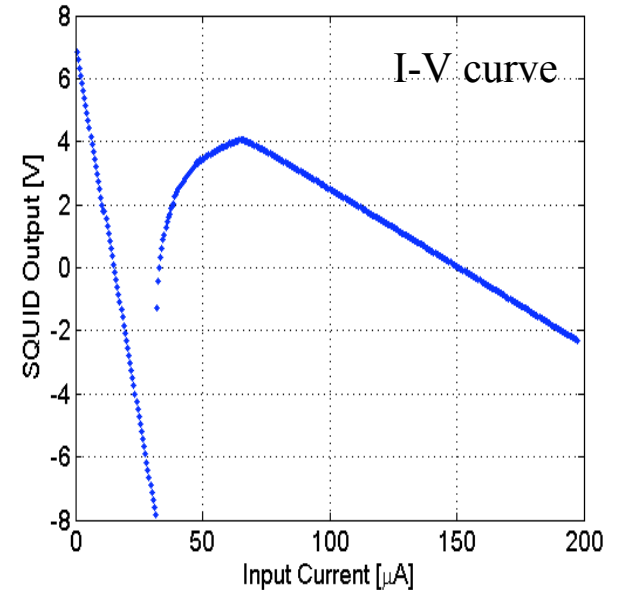
- Tc measurements
- Antenna beam characterization
- Antenna spectrum and band pass filter characterizations
- Optical efficiency of antenna-coupled TES detectors



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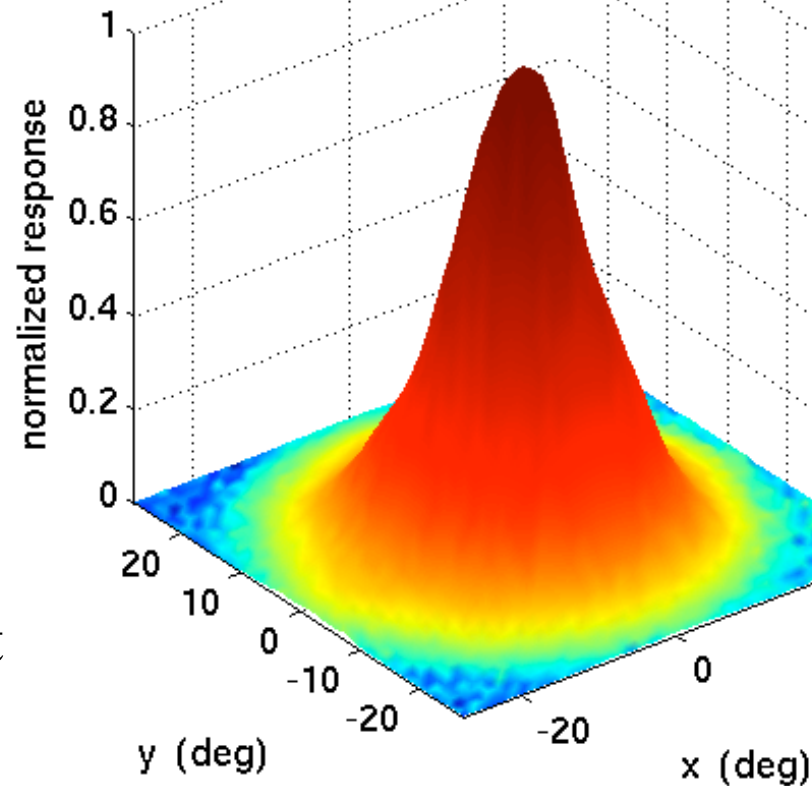


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# Beam of Antenna

- Thermal source
- Two dimensional mapping
- $13.6^\circ \pm 0.1^\circ$   
Gaussian beam at FWHM
- Ellipticity  $< 3\%$
- Cross polarization  $< 5\%$
- Low sidelobes (out of view here)

Prototype 100 GHz  
single pol filtered device

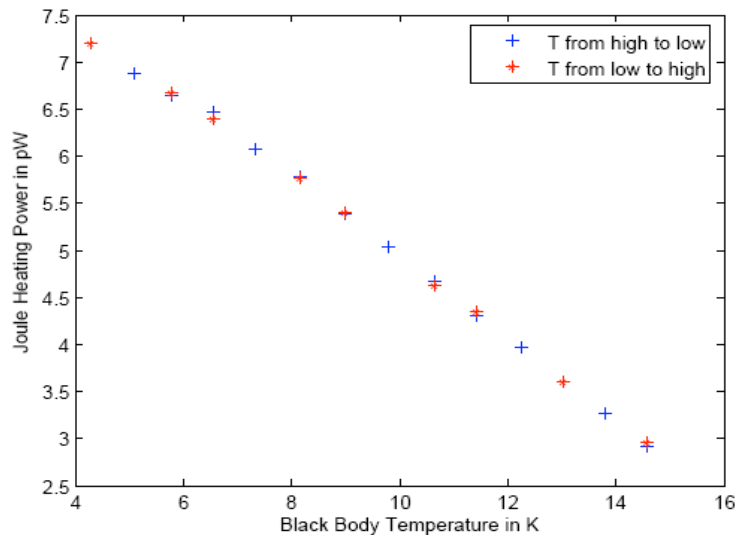
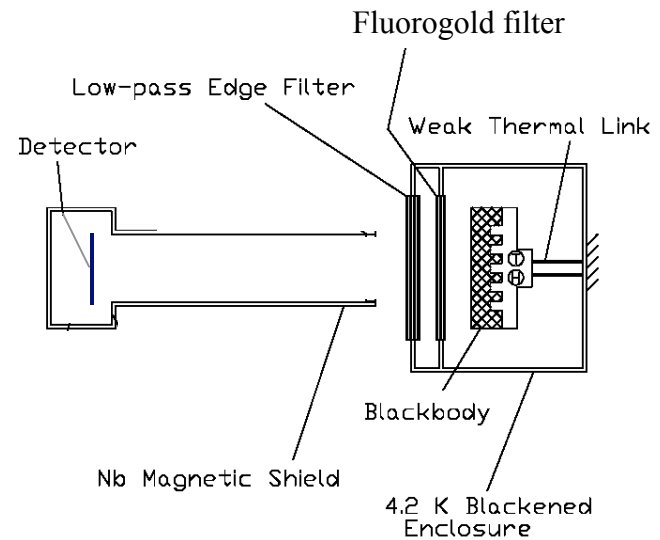
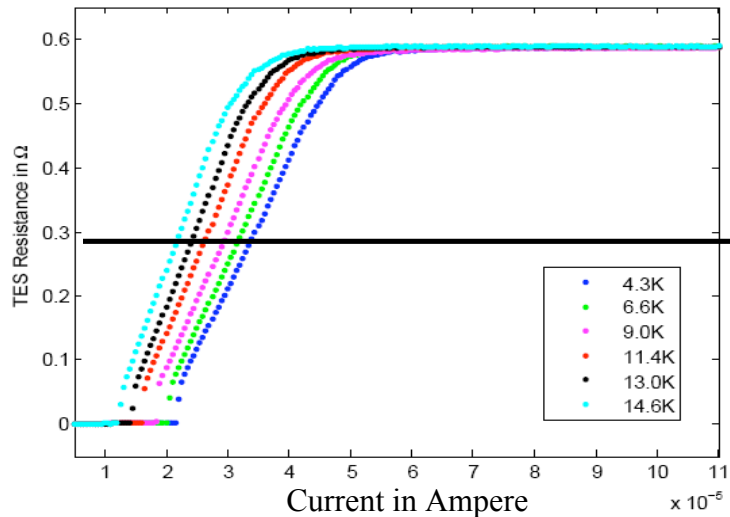


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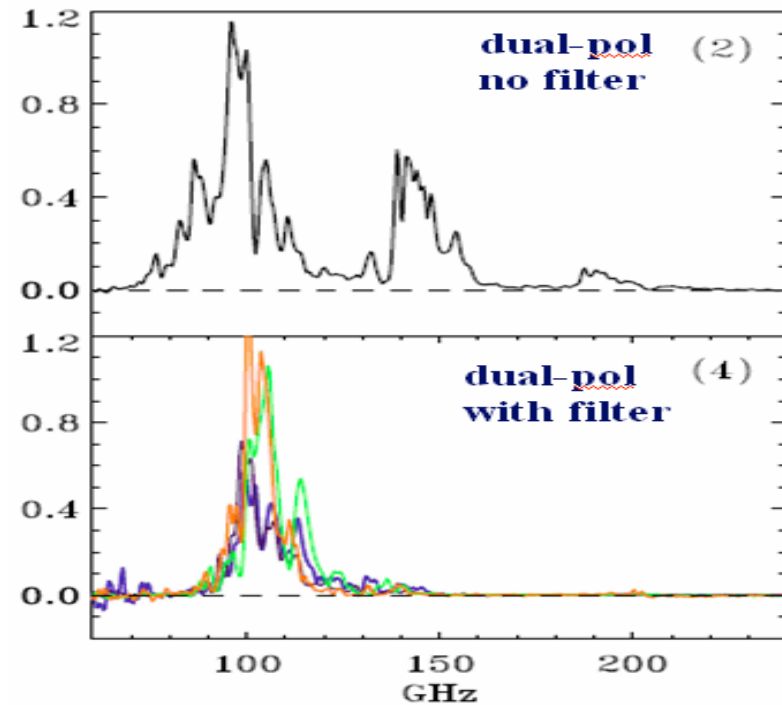
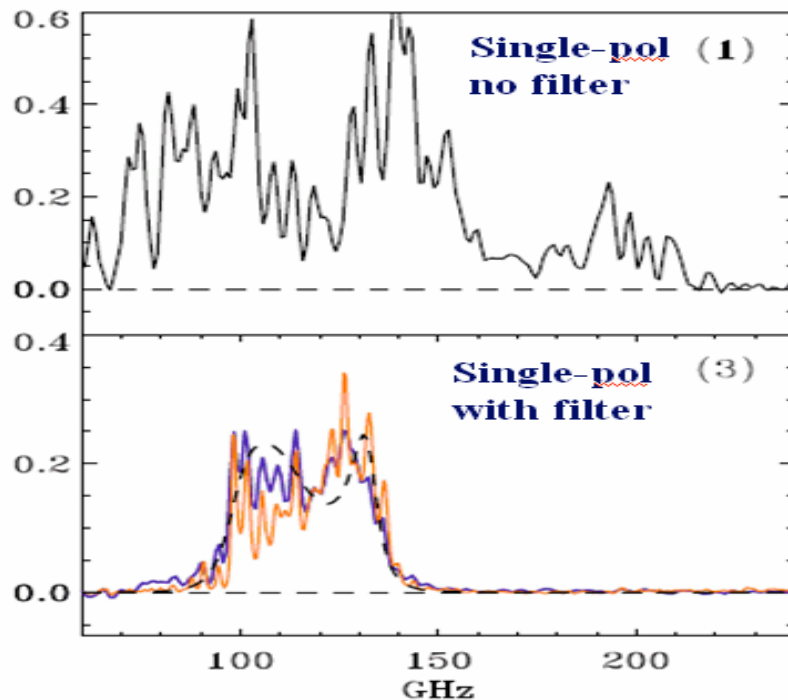
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# Optical Efficiency Test



- Cold blackbody in Dewar
- 4K ~ 20K
- Slope is used to normalize the spectra
- $G \sim 500 \text{ pW/K}$  at 0.9 K

# Spectra

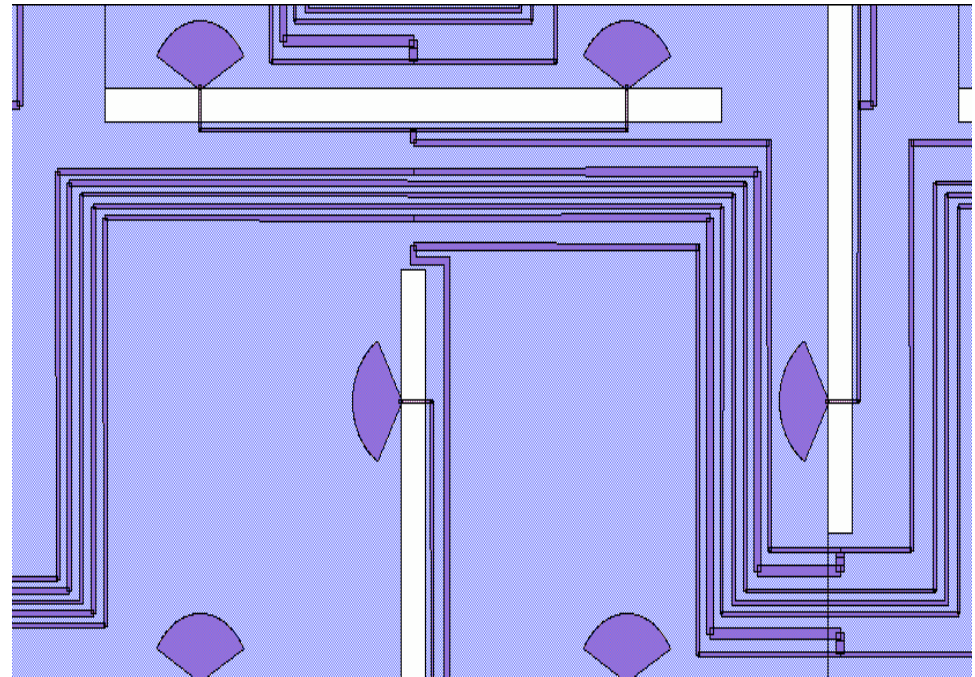
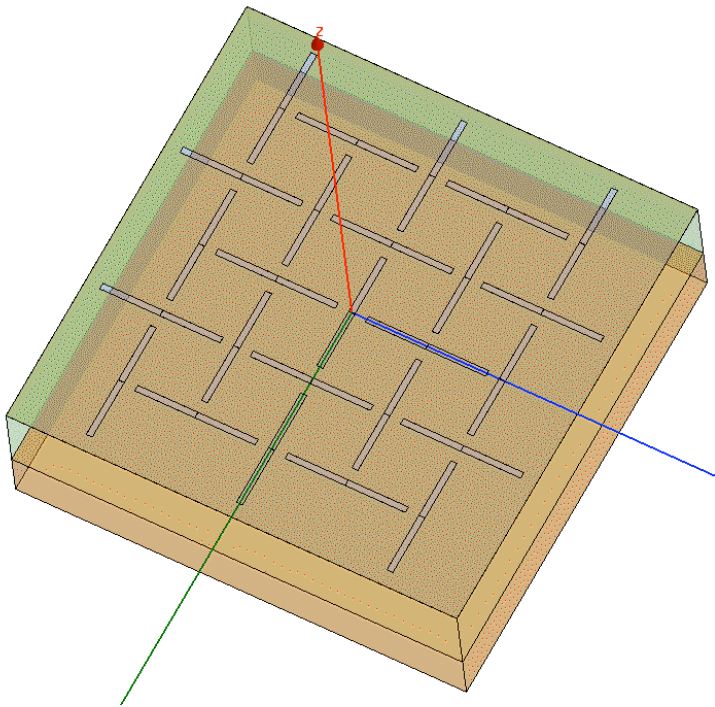


- **Wide band single polarization antenna:** low optical efficiency and fringes caused by feed network
- **Dual polarization antenna has usable bandwidth  $\sim 20\%$ ; high optical efficiency**
- **The lumped element filters produce satisfying band  $\sim 30\%$ , no leaks**



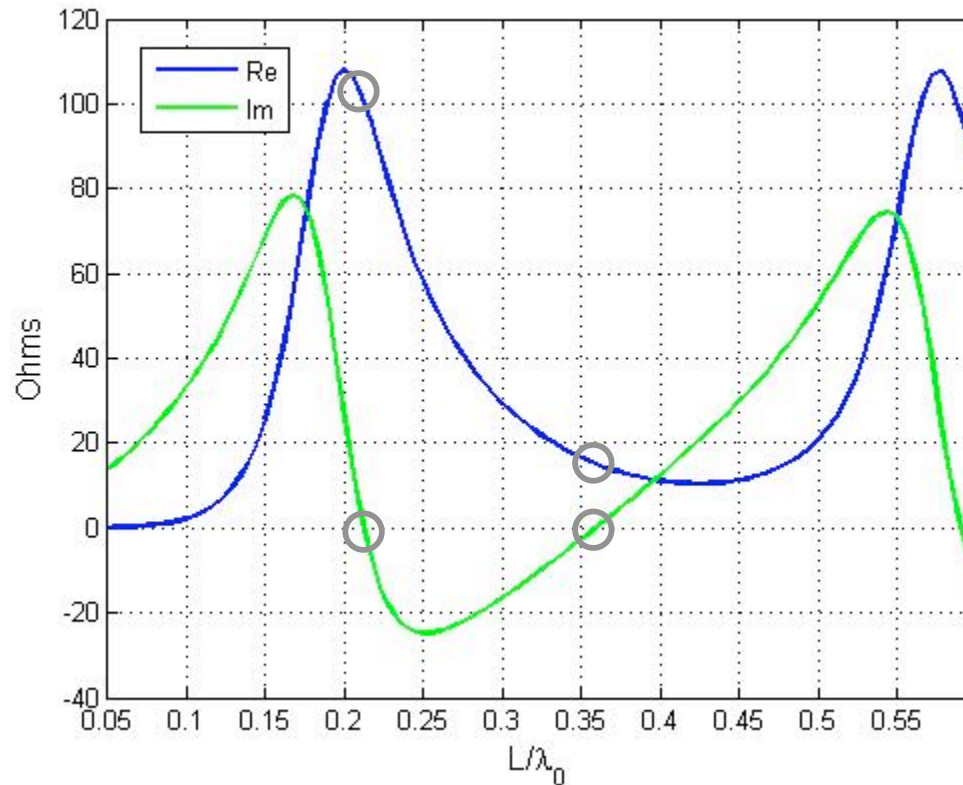


# 145 GHz Dual-Pol Antenna Array



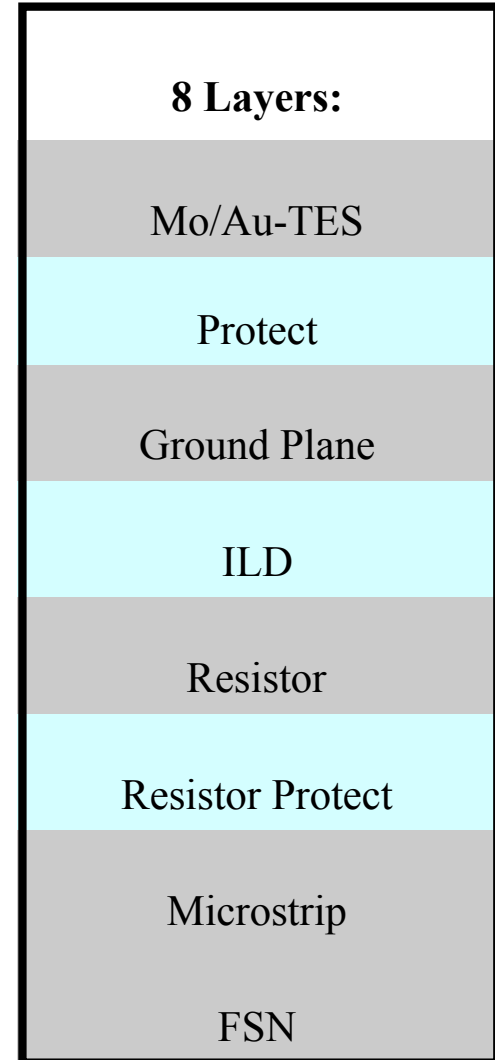
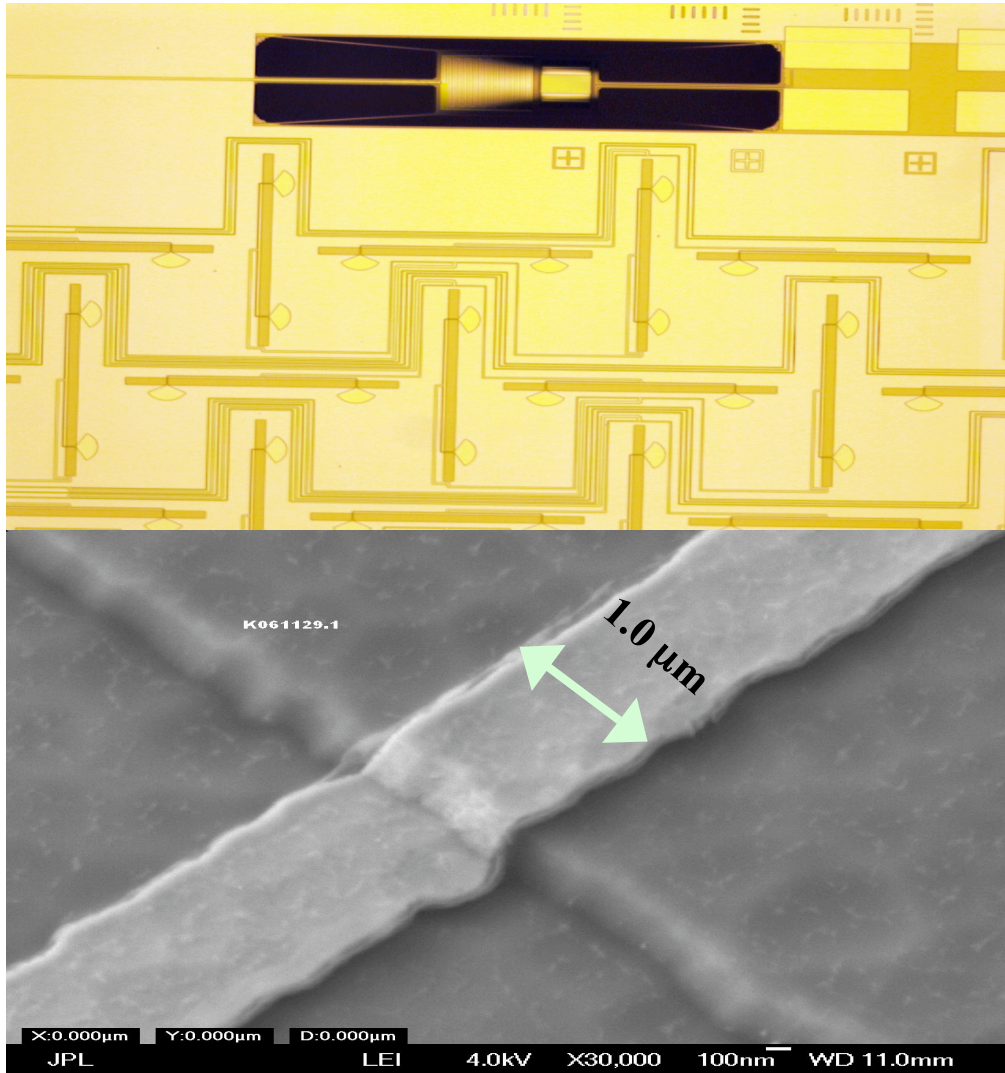
- New design with HFSS simulation (Goutam Chattopadhyay) and moment method calculation (Peter Day)
- **Reduced cross polarization**
- **Increased band width**
- High impedance ( $\sim 40 \Omega$ ), Nb transition line is really narrow

# Antenna Impedance Plot



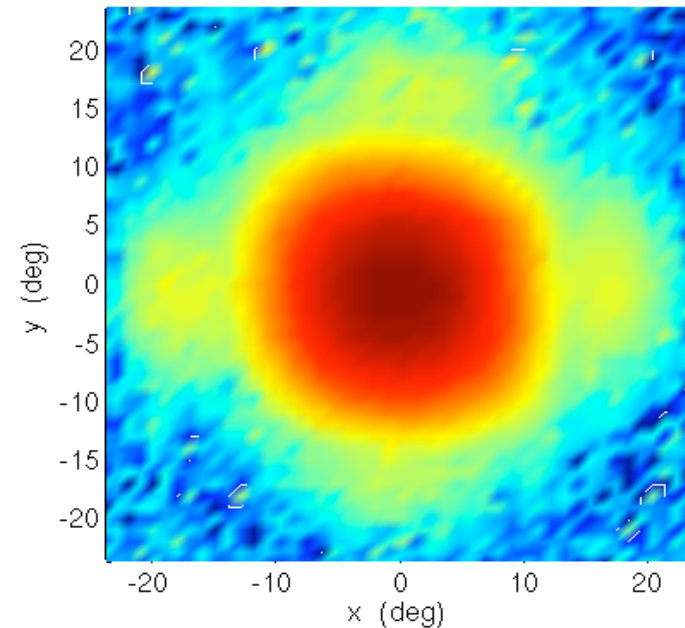
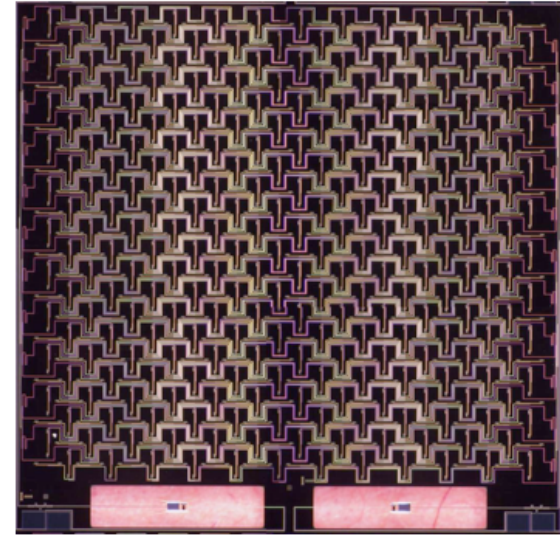
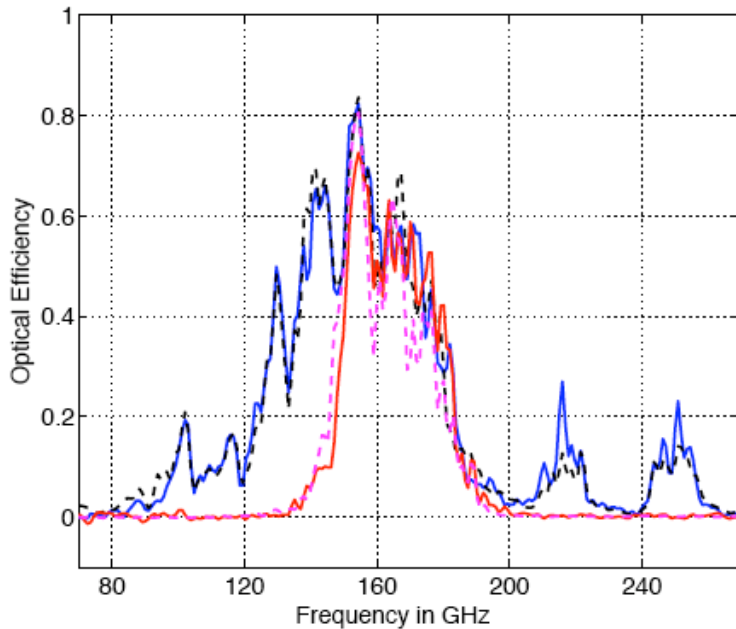
- At the resonance ( $\text{Im } Z = 0$ ), the feed sees a simple resistive load
- The feed transmission line needs to match the impedance of this load
- Short slots are more compact, but could reduce transmission microstrip width to below  $1\mu\text{m}$  due to high impedance of the slot antenna

# New Dual-pol Antenna



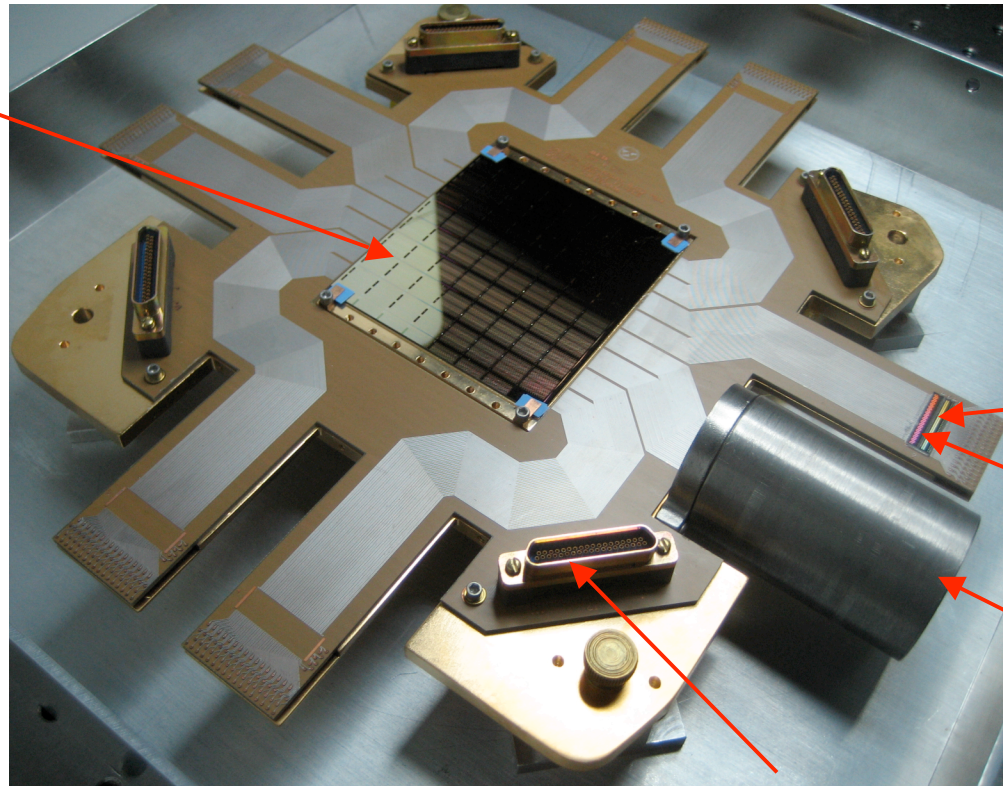
# New Antenna – coupled TES

- Molybdenum gold bi-layer TES, Tc is between 450 mK and 500 mK
- Spectra without filters look good, band width is over 30%
- Filter's band pass is clean
- Optical efficiency is at 60% level
- Cross polarization < 2%
- Nice beam, ellipticity < 3%



# Single Tile Focal Plane for Array Test

Array location with  
 $8 \times 8 = 64$  pixels  
 $8 \times 8 \times 4 = 256$  readouts



Nyquist chip

1st-stage  
SQUID chip

Nb-shield

Connectors for  
-Address lines  
-Detector bias  
-Output to 2<sup>nd</sup>-stage  
SQUIDs

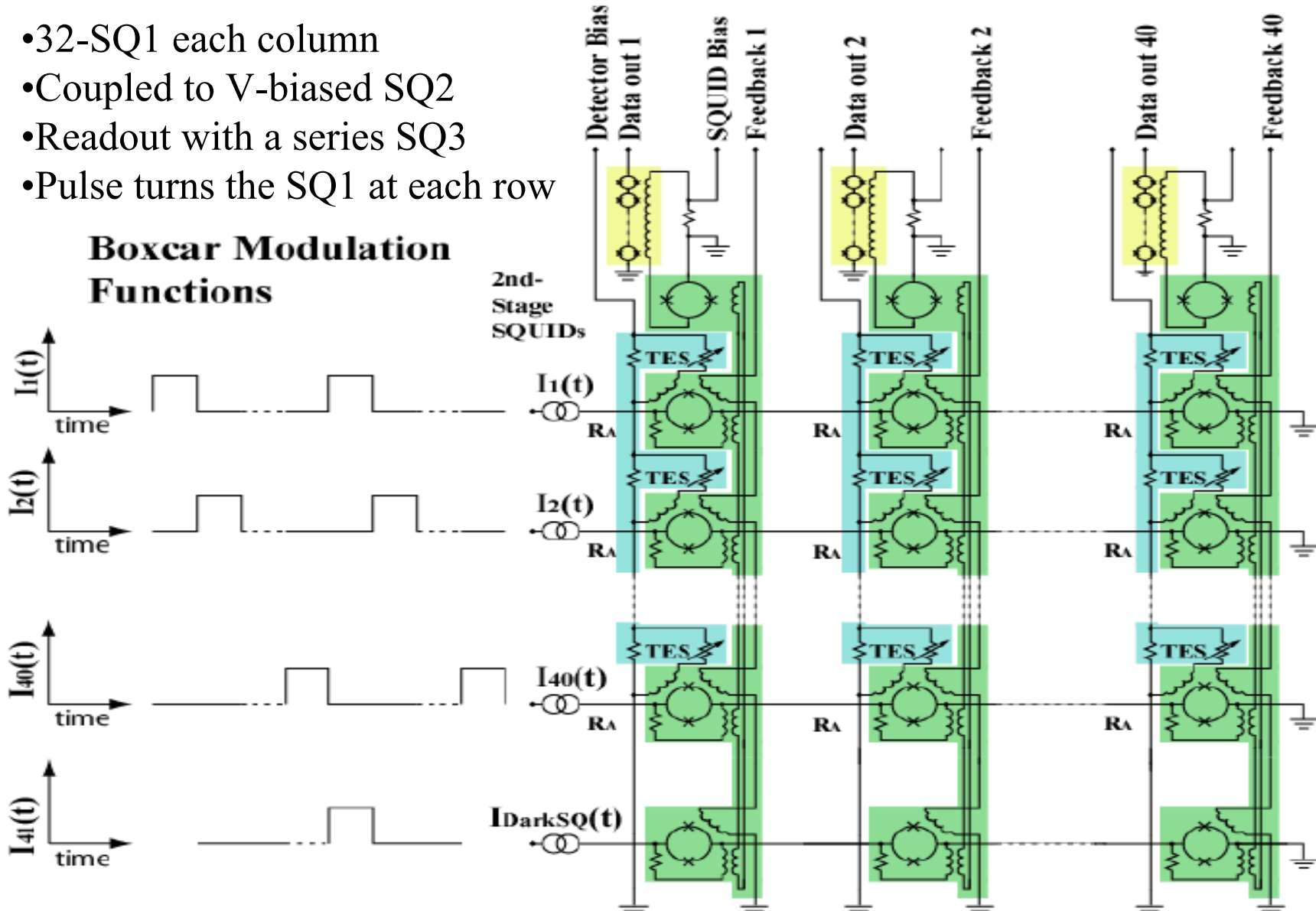
Lab test under the way at JPL,  
limited by cryogenic issues

# Time-Division Multiplexing

De Korte et al., NIST

- 32-SQ1 each column
- Coupled to V-biased SQ2
- Readout with a series SQ3
- Pulse turns the SQ1 at each row

## Boxcar Modulation Functions

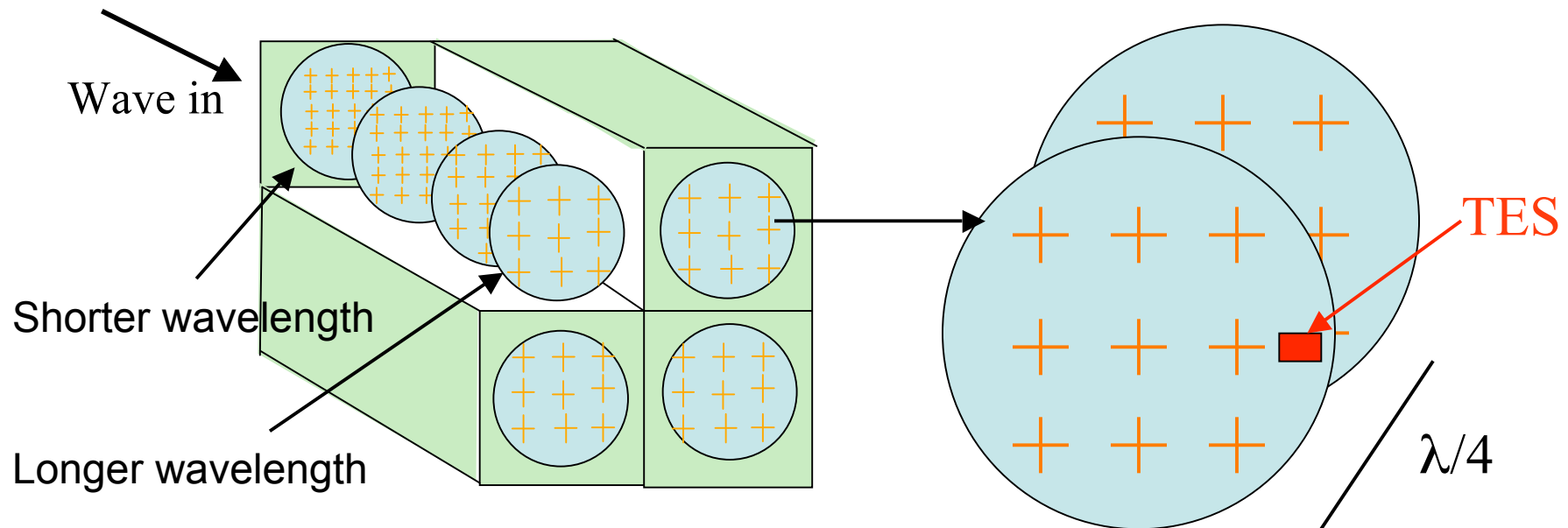


# Excess noise control

- Magnetic field could cause significant noise in TES and SQUID
  - Vortex motion in TES film
  - Trapped flux of DC field in SQUID
  - Trapped flux motion due to strayed AC field in SQUID
- Sources of field: Earth's field, waveplate motor, electronics, etc
- SQUID is very sensitive to magnetic field, to reach  $1\mu\text{K}$  resolution sensitivity, magnetic field need be controlled below  $5\text{nT}$
- Magnetic field effect in TES is about  $1.5\text{mK}/\text{Gauss}$ , TES with voltage bias requires magnetic field below  $1\mu\text{T}$
- Multilayer magnetic shielding: cryoperm + niobium tube
- Excess noise appears at low bias of TES

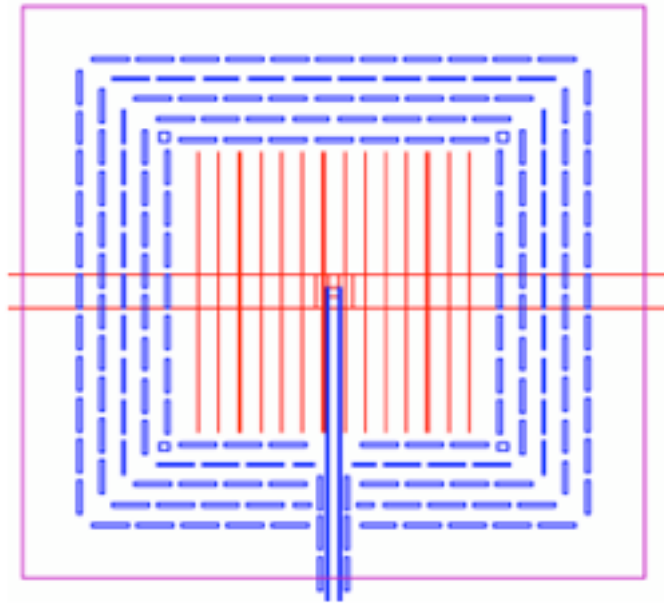
# Multicolor Frequency Selective TES array

- Part of SPEctral Energy Distribution ( SPEED) project to study the spectral energy distribution of high red-shift galaxies. The star formation in the early universe
- 4 x 4 pixel array operating @ sub 0.5 K temperatures; each pixel is a frequency selective bolometers stack that enables sensing radiation at 150, 220, 270, 350GHz
- Simultaneous spatial and spectral measurements
- Efficient use of focal plane area



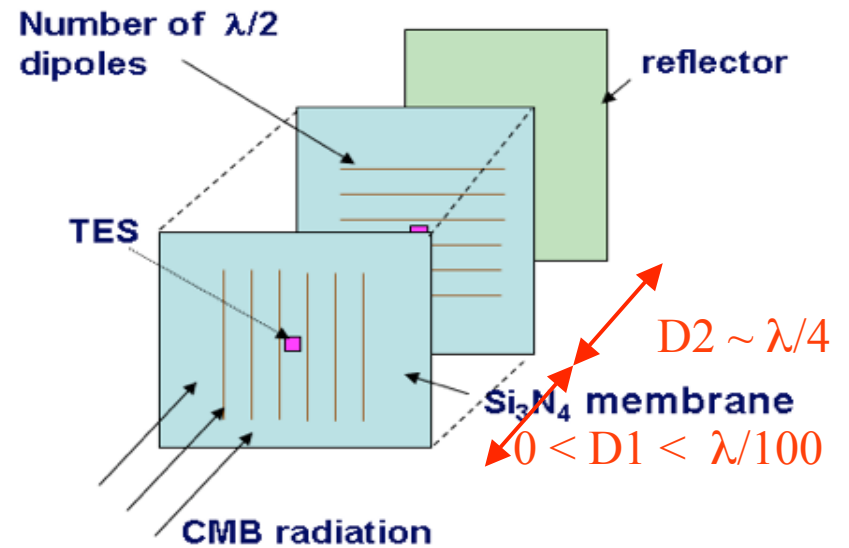
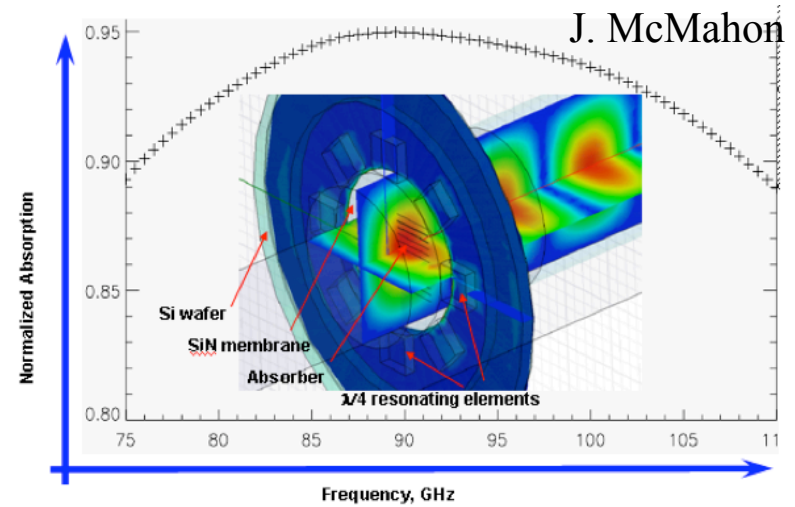


# Absorber Type Polarimeter

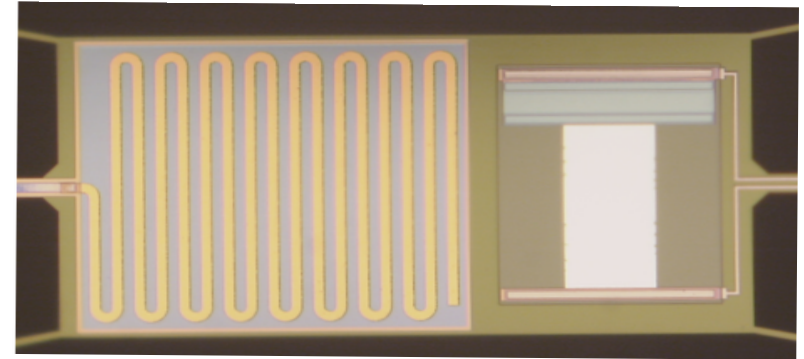
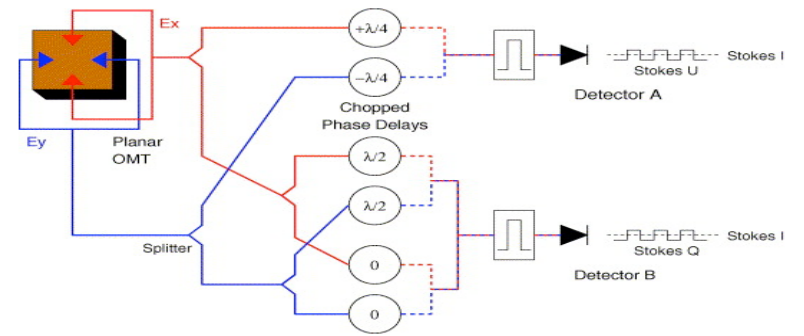
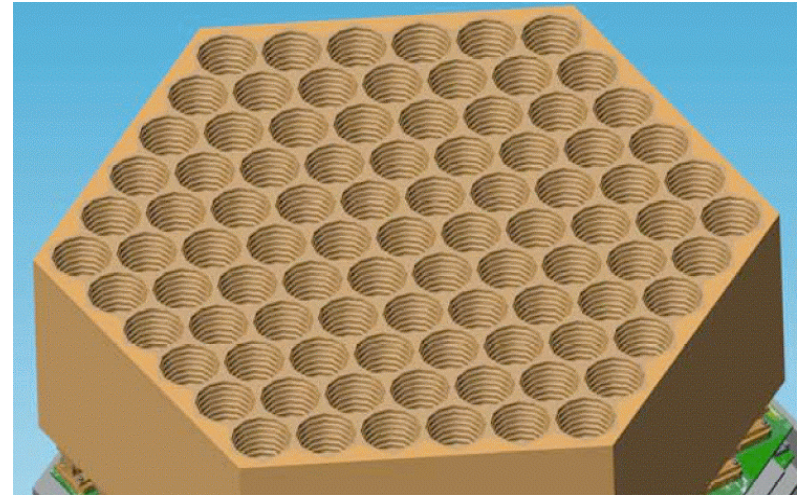
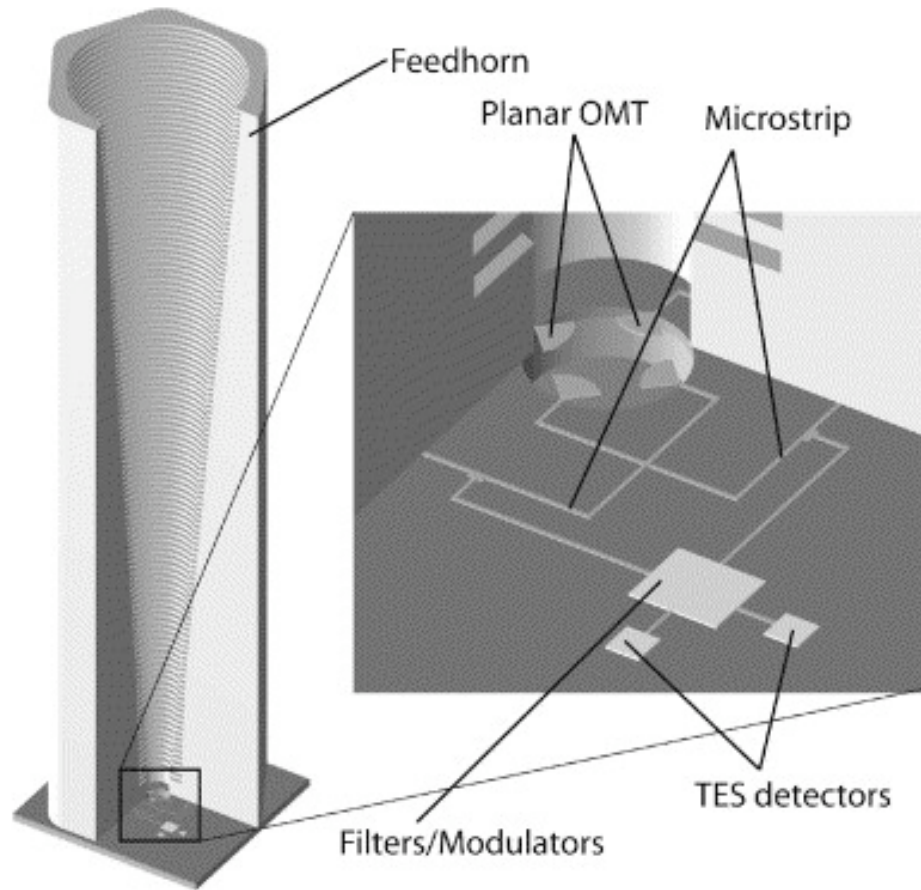


**Simulation for 95GHz**  
 14 stripes, length 1.660mm,  
 width 3 $\mu$ m,  $R=7\Omega/\text{sq}$

$$Z \approx R \frac{L}{W(N+1)}$$



# Integrated large array for Q/U measurements



PAPPA,  $X_{pol} < -45dB$

# Current Status and Future Plan at ANL

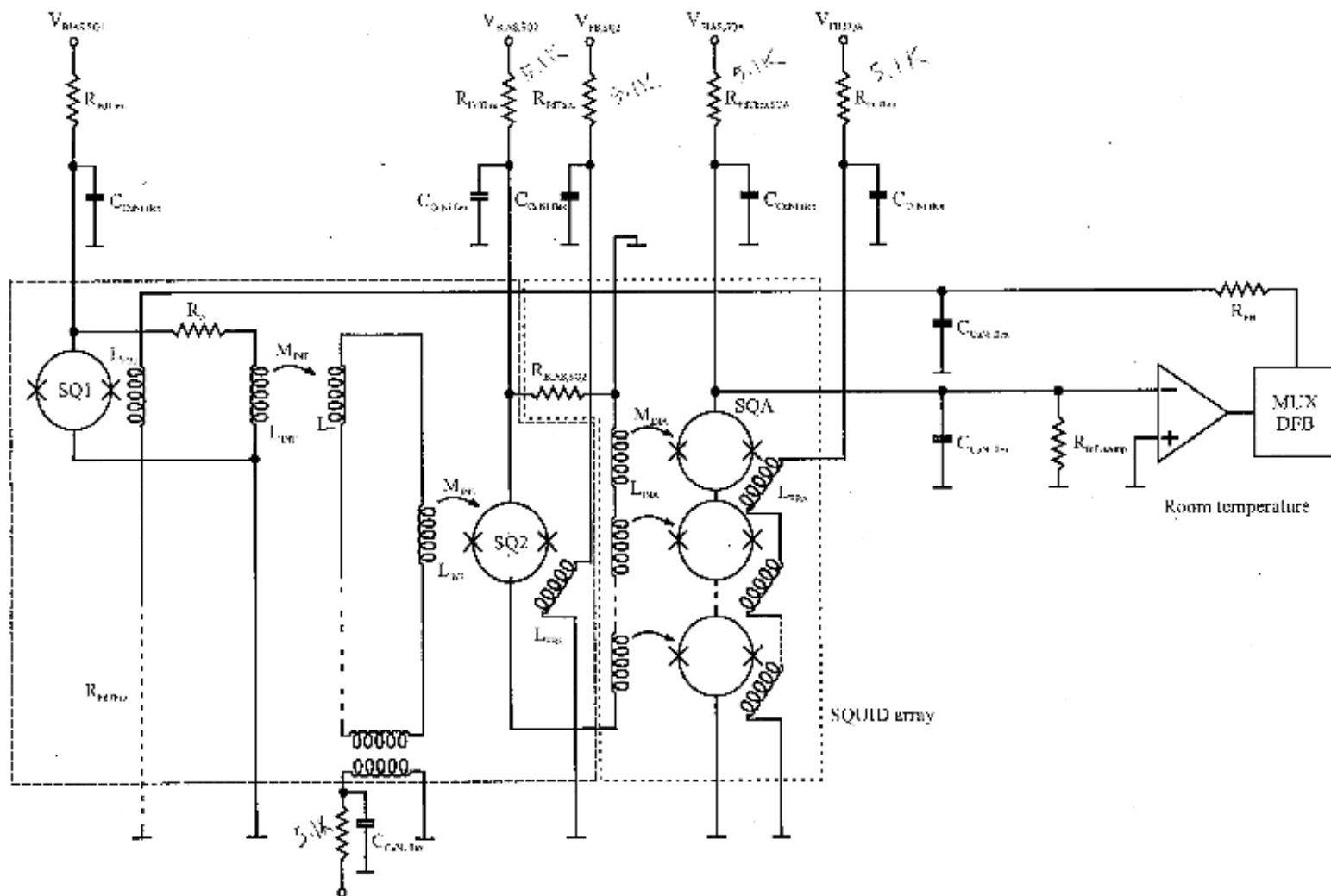
- Frequency selective bolometers for SPEED
  - Micro-fabrication of prototype array;
  - Dark and optical measurements, noise characterization;
  - Device assembling for ground telescope observation
- Polarimeters for SPT
  - Simulation and design (HFSS)
  - Thermal modeling and G tuning
  - Microfabrication of prototype devices
  - Dark and optical measurements
  - Noise characterization
  - Horn coupled large format polarimeters array (thousands of pixels) in polarization-sensitive measurements for the SPT observations
- Enhancing micro-fabrication capability
- SQUID-based multiplexing read-out for detectors array
- New applications, for example, X-ray dispersive fluorescence analysis

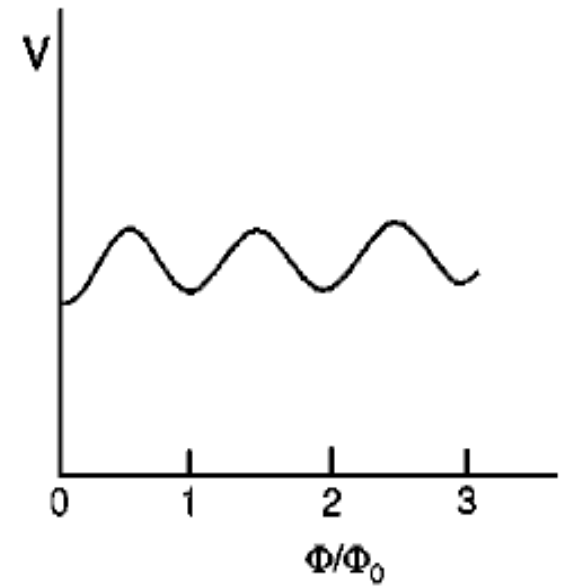
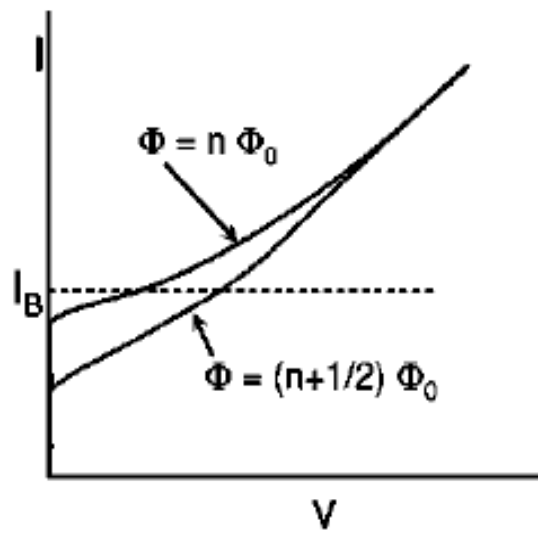
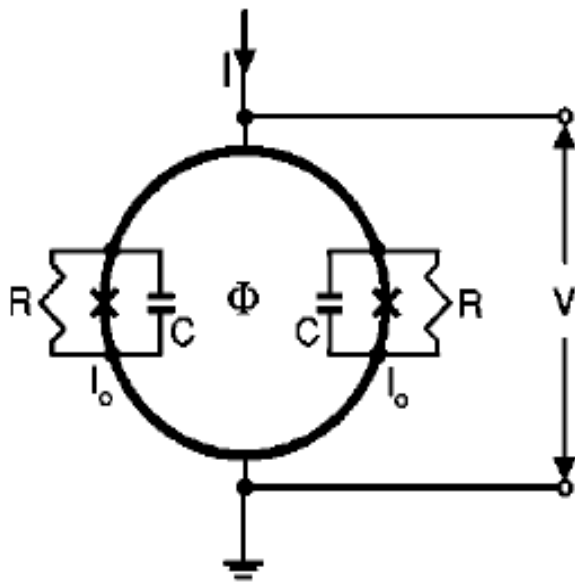
# Summary

- Antenna-coupled TES bolometer technology has been demonstrated
- Large format arrays will instrument a balloon borne CMB experiment -- SPIDER, and an upgraded south pole CMB experiment (BICEP) -- SPUD
- Large format array technology is competitive for future space experiment – CMBpol
- Detector technology efforts at ANL, such as frequency selective bolometers and absorber type polarization sensitive bolometers

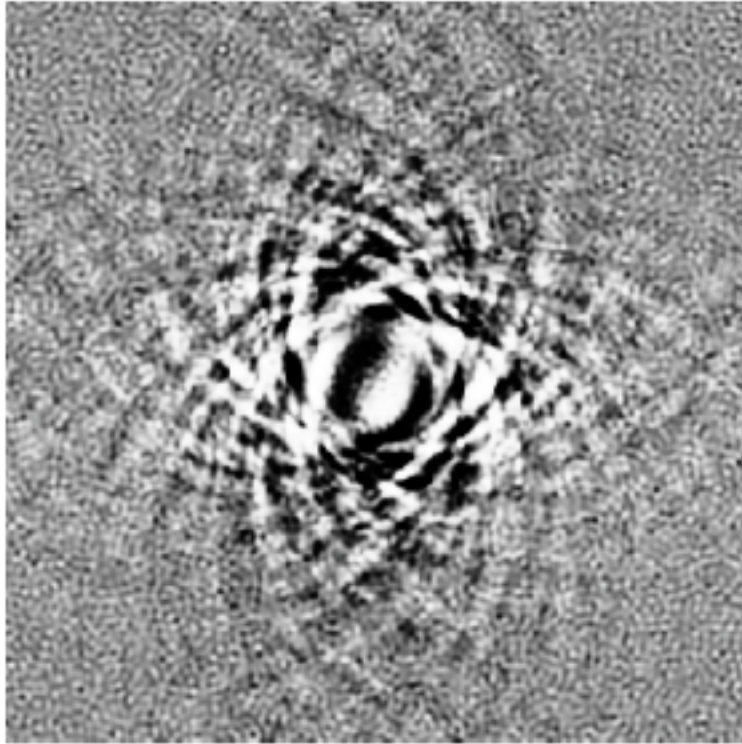
# Acknowledgements

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- **NIST** W. Doriese, G. Hilton, K. Irwin, C. Reintsema
- **ANL** A. Datesman, R. Divan, V. Novosad, J. Pearson, G. Wang, V. Yefremenko
- **U of C** J. E. Carlstrom, C. Cheng, T. Downes, J. McMahon, S. S. Meyer





# B-mode Lensing



W. Hu and T. Okamoto,  
ApJ 2002

$$NEP_{BLIP} = \sqrt{\sum [2P_i(h\bar{\nu} + \varepsilon_i\eta_i k_B T_i)]}$$

The first term is the contribution from photon random arrivals (shot noise due to Poisson statistics). The second term accounts for the effect of photon correlation and depends on the source emissivity and temperature, and the net efficiency through the optics to the detector

$$A = \left( \frac{4n}{(n+1)^2} \right) \frac{4nZ_0R_0}{((n+1)R_0 + Z_0)^2}$$